Research Article
Wideband Negative Permittivity and Double Negative Fishnet-Mushroom-Like Metamaterial in X-Band Waveguide

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1. Introduction

Over the past decade, interest in applications metamaterials (MTMs) has grown explosively; metamaterials are a kind of a material with electromagnetic properties, which are not readily available in nature. Under certain conditions, these materials provide a simultaneous negative permittivity (−ε) and permeability (−μ) in certain frequency range, and they exhibit reversal of Snell’s Law (negative refractive index) and Doppler Effect and Cerenkov Effect around the transmission peak. Depending on the negative value of the permittivity (real part), permeability (real part), or both, metamaterials can be categorized as epsilon negative (ENG), mu negative (MNG), and/or double negative (DNG), which presented the left-handed metamaterials (LHM) they were first proposed by Veselago [1]. Conventional materials have permittivity and also permeability larger than zero in all frequency ranges and they are called double positive (DPS). They become very desirable for designing and fabricating MTMs that can achieve these properties, according to their possible applications in diverse engineering and scientific fields. The metamaterials are designed to use dielectric as host and metal containment, which have been fabricated and then tested intensively in the form of different periodical arrangement [2, 3]. Recently, the fishnet structure has attracted a great potential approved in the studies where the negative refraction index for first time was displayed in a 3D optical MTM [4, 5]. Basically, the strategy to acquire the negative permeability is to create exciting circular currents that can generate a strong magnetic resonance, while being expected to produce the negative permittivity by virtue of the electric plasma response with added continuous wires. When negative refraction index is inconsistent with the negative permeability, most of fishnet structure researches concentrated on characterizing and modeling their magnetic resonance [6–9].

The dispersion diagram of a 2D-periodic structure is produced by calculating the resonant frequencies of Eigen modes, which present differential equation solutions that simulate the fields in a single MTM cell. However, a different band gap acquired by a restricted amount of surface waves along the contour of the irreducible Brillouin zone for an infinite array of the proposed MTM structure will be

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produced depending on how many metamaterials arrays are presented. Therefore, only the dispersion relation can obtain the exact position of passband and band gap in the frequency spectrum.

The study presents two cases of MTMs. The first one is a wideband \((-\varepsilon)\) negative permittivity (ENG), whereas the second one is double negative (DNG). Both cases are the electromagnetic behaviour represented in \(S\)-parameters that are numerically and experimentally studied. The simulated and the experimental results exhibit very good agreement in the (8–12) GHz X-band regime whereas the permittivity and permeability and refraction index characteristics are extracted to explain the behaviour of the MTM cell. The proposed MTM (Fishnet-Mushroom-like) is fabricated and measured. The obtained results are compared with simulated results, which are carried out by CSTMWs' simulation program [10]. The given notes, as well as the Nicolson–Ross–Weir (NRW) procedure, are considered to obtain the properties of the designed MTM structure and the effective material parameters [11, 12].

2. Features of Electromagnetic Band Gap (EBG) Structure

2.1. Unit Cell Design. The proposed metamaterial EBG unit cell consisted of a square substrate with 4.5 mm side length and 1.6 mm thickness. The FR-4 epoxy with a permittivity \(\varepsilon_r = 4.3\) and loss tangent \(\delta = (0.025)\) is chosen to represent the substrate material. The metal shape is printed on the substrate front side which is composed of a squared patch surrounded by two square rings connected from their center sides like a net. A cylindrical via with a diameter of 0.5 mm is also located in the unit cell center as shown in Figure 1(b), by which the square patch and the ground plan in the bottom side are connected. Copper of electrical conductivity \(\sigma = 5.8 \times 10^7\) S/m with coating thickness 0.017 mm is chosen for printing as shown in Figure 1(a). The unit cell dimension parameters are mentioned in Table 1.

<table>
<thead>
<tr>
<th>Parameter</th>
<th>mm</th>
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<tbody>
<tr>
<td>(a)</td>
<td>2.4</td>
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<tr>
<td>(b)</td>
<td>0.2</td>
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<tr>
<td>(c)</td>
<td>4.5</td>
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<tr>
<td>(d)</td>
<td>0.5</td>
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<td>(h)</td>
<td>1.6</td>
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2.2. Dispersion Diagram. The concept of this paper states that the wave propagation direction is assumed in the same direction of the unit cells periodicity. The space harmonics share the same group velocity even if they have different phase velocities for they cannot occur separately. Each single harmonic does not satisfy the boundary conditions of the periodic structure, but their summation does so. Thus, summation is considered to be the same mode. Hence, by simulating a single proposed EBG unit cell structure, an infinite periodic structure model is carried out by CSTMWs eigenmode simulator.

The band gap characterization for EBG is shown in Figure 2. We expound the idea that the field for the first mode is TM dominant, started at zero frequency, and the eigenfrequency increases with the wave number, reaching the maximum frequency 8.45 GHz. It also decreases following the light line down to a certain frequency. The second mode starts at a higher frequency and continues increasing with a slope under the vacuum's speed of light, which is set by a straight line. The band gap is near 4 GHz for this particular structure (8.45–13.15 GHz), comparing to previous structure designs that achieved less band gap in the same arena [13, 14].

2.3. Array Structure Design and Simulations Settings. The simulations included the proposed unit cell and are ordered by varying different arrays arrangements [15], like \(1 \times 8\)-cell array and \(3 \times 8\)-, \(4 \times 8\)-, and \(6 \times 8\)-cell arrays. These
3. Results and Discussion

For an experimental reason to suit with the WR-90 dimensions and to decrease the misalignment experiment duration such as position of the fabricated sample plate, the array of $4 \times 8$ arrangement is chosen and fabricated as shown in Figure 4(a), whereas Figure 4(b) showed the schematic experiment.

The reflection coefficient $S_{11}$, transmission coefficient $S_{21}$, and the reiterative characteristic parameters have a negative index, whereas the permittivity and the permeability are in the X-band frequency range. The simulation results exhibit that the structure has been designed successfully; that is, the simulation and experimental data agree with each other as possible.

The experiment steps are as follows: First, the WR-90 waveguide is connected with the input and the output ports with the network analyzer by the two coaxial probes, and the network analyzer is calibrated. Second, the fabricated sample is fixed at the center of a WR-90 waveguide; all the metallic part that is used is unconnected to the waveguide inner walls. The dielectric substrate thickness has chosen being as thin as possible to minimize the dielectric loss. The selected structure dimensions which severed the optimization goal had proven the double negative performance at around X-band arena [11, 12].

arrangements are set between two waveguide ports in the $x$-directions in order to simulate $S$-parameters and obtain the reiterative characteristic parameters, and the band gap effect of varying the amount of metamaterial is so.

The EBG is analyzed and simulated by applying the sample to a surface wave propagation measurement through the setting procedures that presented in (Figures 3(a) and 3(b)).

Furthermore, microwave ports are set on the $x$-axis sides, on which one port is located as a transmitter generating electromagnetic wave ($E^+$, $H^-$) and the second port as a receiver. These boundary conditions are chosen in order to define a waveguide structure (Figure 3(b)). Moreover, a perfect magnetic boundary condition (PMC) is set in the $y$-directions in which infinite periodic repeating in the same

directions for the eight EBG and a perfect electric boundary condition (PEC) are set in the $z$-directions, considering a ground plane placed in the below side of the EBG structure and with the open boundaries in the $x$-directions where the ports are set.
Referring to the method described in Section 2.2, surface waves propagating and travelling through the circuit board are detected and the amplitude and phase for the reflection and transmission data measured. Figure 5(a) shows both simulated and measured S-parameter data for the investigated structure in good agreement between them.

The experimental and the simulated results show a small difference in minor deviations between the measured and the simulated data that are pointed to manufacture allowance, which is related to PCB fabrication and the gathering process and the quality of the dielectric FR4 dispersion used as a substrate. Likewise, the misalignment experiment duration such as position of the fabricated sample plate, which cannot be neglected, port receptacle locations, and loss characteristics of the materials are regarded as another root of error. On the other hand, the extent of measurement accuracy can be elucidated through the convergence between the results of experimental and simulations outcome.

The effective parameters for the obtained investigated structure material are based on the effectively homogeneous medium occurring when the periodicity of the structure is much less than the propagation wavelength in the propagation direction. The experimented metamaterial is treated as a LHM in the region between 10.95 and 11 GHz, and the transmission peak is indicated at 10.975 GHz. It is clearly stated that the boundary conditions selected to consider the designed structure as an infinite homogeneous material in a certain region led to a frequency observation [14], where both the dielectric permittivity and magnetic permeability are simultaneously negative, which resulted in the retrieval method application.

The NRW method is used in the material parameters calculation extracted and presented in Figures 5(a) and 5(b). As the figure shows, the metamaterial structure exhibits a negative permittivity in all X-band simultaneously with negative permeability in a band between 10.950 and 11 GHz, which a high band transmission peak covered. In spite of the negative refraction index, wideband included all the X-band (see Figure 5(b)), and that type of the negative refraction index is not supporting the left-hand (LH) behaviour. As it was previously mentioned, the representation of the single negative (SNG) occurred only with negative real part of the permittivity. Consequently, both the single negative (SNG) (between 8 and 10.95 GHz and between 11 and 12 GHz) and the double negative (DNG) (10.95–11 GHz) LH region characteristics have been supported by the real part of the negative refractive index [16, 17]. Due to the electric field and the copper parts conjunction and because of strips as in the continuous-wires case, the permittivity exhibits a Drude-like behaviour. Also, the permeability shows Lorentz-like behaviour, through a strong resonant response to the magnetic field during the magnetic resonators condition.

Figures 6(a) and 6(b) show the simulated propagation of surface waves and currents due to the presence of EBG 4 × 8 structure. It is exhibited from these simulations, when surface currents are generated from (port 1) side in x-direction which then travels through the EBG structure opposite side (port 2). In Figure 6(a), the EBG structure with an operating frequency band —the single negative (SNG) (between 8 and 10.95 GHz and between 11 and 12 GHz)—was applied. As expected, the band gap effect shown in Figure 2 does not support any wave propagation at a frequency included within its band gap and suffering reduction. On the other hand, Figure 6(b) demonstrates the LHM in the region between 10.95 and 11 GHz and as a consequence of LHM behaviour the EBG structure surface presence results in a symmetric distribution of the surface current from port 1 to port 2.
4. Conclusion

In this study, the proposed Fishnet-Mushroom-like MTM is fabricated and measured, and the obtained results are compared with simulated results. There are two cases of MTMs, the first one of which is \((-\varepsilon)\) a wideband negative permittivity (between 8 and 10.95 GHz and between 11 and 12 GHz) and the second one is double negative (DNG) (between 10.95 and 11 GHz, LH region; and 10.975 GHz is indicated as transmission peak). The electromagnetic behaviour for the said cases in question is represented in \(S\)-parameters numerically and experimentally for the X-band arena. The value of this study brings forth the fact that this structure can be used in wideband (ENG) applications such as efficient electrically small antenna enhancement applications and also for double-negativity antenna and radar cross section (RCS) property enhancement applications [18–21].

Competing Interests

The authors declare that there is no conflict of interests regarding the publication of this paper.
Figure 6: (a) Showing the simulated propagation of surface currents for 10 GHz in the SNG region. (b) Showing the simulated propagation of surface currents for 9.975 GHz in the DNG (LHM) region.

References


