

## Research Article

# Payload Parameter Identification of a Flexible Space Manipulator System via Complex Eigenvalue Estimation

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Manipulator systems are widely used in payload capture and movement in the ground/space operation due to their dexterous manipulation capability. In this study, a method for identifying the payload parameters of a flexible space manipulator using the estimated system of complex eigenvalue matrix is proposed. The original nonlinear dynamic model of the manipulator is linearized at a selected working point. Subsequently, the system state-space model and corresponding complex eigenvalue parameters are determined by the observer/Kalman filter identification algorithm using the torque input signal of the motor and the vibration output signals of the link. Therefore, the inertia parameters of the payload, that is, the mass and the moment of inertia, can be derived from the identified complex eigenvalue system and mode shapes by solving a least-squares problem. In numerical simulations, the proposed parameter identification method is implemented and compared with the classical recursive least-squares and affine projection sign algorithms. Numerical results demonstrate that the proposed method can effectively estimate the payload parameters with satisfactory accuracy.

## 1. Introduction

Manipulator systems are an important component of the spacecraft structure. These systems are currently widely applied to the on-orbit servicing missions, such as space object capture and structure assembly [1–3]. Manipulator systems in space are generally designed as long span and lightweight to increase workspace and reduce launch mass. Thus, the flexibility of links is evident. When the manipulator carries a large endpoint payload or captures a space object, the vibration of the system is obvious, even with only a slow movement. For example, the operators of the shuttle remote manipulator system (SRMS) spend 20 to 40 s waiting for oscillation decay after maneuvering the arm to avoid the vibration problem in an on-orbit operation [4]. In the operation of the flexible space manipulator system, guaranteeing an accurate positioning in pick-and-place tasks is a crucial aspect that must be addressed. One of the major obstacles

to overcome is designing a control method capable of canceling the vibrations when the system dynamic model is affected by the changes in the payload parameter. If these changes are disregarded in the controller design, then the algorithm may lose accuracy and effectiveness in the vibration suppression, and the system may become unstable in some cases.

Therefore, the payload parameters are identified to obtain the new system characteristic after the operation. The identified parameters can be used to adjust the adaptive control law and provide an important reference for improving the overall performance of the system [5–7]. In particular, if the manipulator is used to capture a payload with an unknown property (e.g., a noncooperative object), then accurately identifying object parameters can be implemented to recalculate uncertain parameters periodically and update the certainty equivalence controller with unexpected parameter variations conveniently [8]. Payload object capture and corresponding parameter identification experiments using

the space manipulator have been successfully implemented in certain on-orbit spacecraft, such as the Engineering Test Satellite-VII (ETS-VII) [9], Japanese Experiment Module Remote Manipulator System (JEMRMS) [10], and Orbital Express [11]. However, these experiments primarily focus on cooperative target capture. Thus, the payload parameters are identified, and all manipulators involved are considered rigid structures. Therefore, certain problems regarding structural flexibility and unknown payload parameters in the manipulator system require further study.

Various identification methods have been proposed to estimate the inertia parameters (i.e., mass, mass center, and moment of inertia) of the manipulator and the unknown payload. Most of these methods are least-squares (LS) techniques, which are validated by JEMRMS experiments using space manipulators [9, 12–14]. Based on the planning for optimal maneuvers, the inertia parameters of the ETS-VII spacecraft and the manipulator payload are obtained by solving a nonlinear LS problem [9]. However, alternative methods, such as the linear and angular momentum conservation [15, 16], Kalman filtering [17], and the algebraic identification approach [18], are available. Nguyen-Huynh and Sharf developed a new inertia parameter identification method based on the momentum conservation equation and recursive least-squares (RLS) estimation after the space manipulator grasped an unknown tumbling target [15]. Bruggemann et al. also proposed a robotic-based identification method of unknown spacecraft inertia properties using the momentum conservation [19]. Liu et al. introduced a recursive differential evolution algorithm to identify the inertia parameters of an unknown target and revise the friction parameters of the space manipulator joints simultaneously [20]. San-Millan et al. developed a technique for real-time identification of the payload mass parameter for a single-link flexible manipulator based on the algebraic identification algorithm [21]. However, several of these studies only considered the manipulator to be a completely rigid structure and neglected the influence of flexible vibration [9, 15, 20]. Furthermore, many inertia parameter identification approaches may be unsuitable for the space structure because the available system excitation and measurement signals are finite in the space environment. Therefore, developing a general on-orbit identification algorithm is necessary.

Notably, the system model and its corresponding eigenvalue parameters may be changed due to variations of the manipulator configuration and payload mass to determine the payload parameters of the flexible space manipulator. Thus, the corresponding system identification issue becomes sophisticated. In this study, only the situation for a flexible single-link space manipulator in a selected working point is considered to simplify the analysis process. Then, the original nonlinear dynamic manipulator model is degraded as a linear model by local linearization technology. Consequently, if the system eigenvalue parameters can be obtained using certain identification algorithms, then the corresponding relationship between the payload and the system eigenvalue parameters can be established from the linearized dynamic equation, and the payload

parameter can be determined from the identified structural model parameters.

In the existing time-domain identification approaches for the model parameters, the eigensystem realization algorithm (ERA) by Juang and Pappa [22] has been successfully applied for parameter identification experiments and operational modal analysis repeatedly [23, 24]. The ERA is a typical system realization method. By constructing the system Hankel matrix in the discrete-time domain, the singular value decomposition (SVD) is used to determine the system order and separate the noise signals. Then, the modal parameters are obtained from the identified state-space model using system impulse responses. However, the slow decaying response for lightly damped systems may produce a large Hankel matrix and long computation time. Therefore, a type of extended ERA called the observer/Kalman filter identification (OKID) method was developed to solve the aforementioned problem using system input and output (I–O) signals simultaneously [25]. The OKID method is conducted by adding an observer into the system and placing the desired eigenvalues. Thus, the observer Markov parameters are dead-beat. The ERA–OKID series algorithms were proven to be reliable identification algorithms with a certain degree of noise immunity due to SVD technology. These methods have already been used for the identification of the state-space model and modal parameters of the ground structure and spacecraft [22, 26–28]. However, studies involving the identification of manipulator parameters using the ERA series algorithm are limited.

The current study mainly is aimed at investigating the payload parameter identification of a flexible manipulator system. A novel method is proposed to determine the unknown payload inertia parameter using the complex eigenvalue estimation of the system. Different from the frequently used inertia parameter identification methods that regard the manipulator as a rigid body [9, 15, 20], the manipulator vibration due to link flexibility is studied, and the nonlinear dynamic equation is established. Subsequently, the nonlinear dynamic equation is linearized at a selected working point, and then the corresponding state-space model is obtained. Moreover, using the designed I–O signal system, the complex eigenvalue parameters of the system are determined by the OKID algorithm. Thus, the payload parameters are derived using the LS method based on the identified eigenvalues and system complex modal analysis. Through comparison with the classical RLS method and the affine projection sign algorithm (APSA), the numerical simulation results demonstrate that the proposed method can be used for payload parameter identification of flexible space manipulator with satisfactory identification accuracy. In practical applications, the proposed method can provide a reference for the capturing process of the unknown object, and the manipulator structural modal and the endpoint payload parameters at the working point can be identified simultaneously.

The remainder contents of this paper are organized as follows. The dynamic model of the flexible single-link manipulator is established and further be linearized at the selected working point in Section 2. In Section 3, the OKID

algorithm is briefly reviewed, and the identification of the payload parameter is studied based on complex eigenvalue estimation. Section 4 is dedicated to the analysis of the numerical simulations, and the identification results of payload parameters using the proposed method are compared with those of the classical RLS and APSA algorithm. Some conclusions are presented in Section 5.

## 2. Model Description of Space Manipulator

In this section, the nonlinear dynamic equation of a flexible single-link manipulator system will be established using the Lagrange method, and then the linearized dynamic model is obtained by the local linearization technology.

*2.1. Dynamic Modeling Based on Lagrange Method.* Before establishing the dynamic model of the flexible single-link manipulator, some assumptions are introduced as follows:

- (1) The manipulator moves in a plane, and the gravity influence is ignored
- (2) The flexible link complies with the Euler-Bernoulli beam theory, and the axial deformations as well as nonlinear geometric effect due to bending are negligible
- (3) The link's cross-sectional area remains constant along the link and the material is homogeneous, so the linear density and Young's modulus of the link are constant
- (4) In the paper, the payload is simplified as a mass point. Therefore, only consider the identification problem of the mass and moment of inertia parameters of the payload
- (5) The flexible of the joint and the system damping are negligible in this paper

Based on these assumptions, the manipulator coordinate system and corresponding parameters are defined as follows: the origin  $O$  of the inertial coordinate frame  $O$ - $XYZ$  and the origin  $O_1$  of the link coordinate frames  $O_1$ - $x_1y_1z_1$  are both selected at the joint. The rotation of the link coordinate frame with respect to the inertial coordinate frame  $O$ - $XYZ$  is defined as  $\theta$ . In addition, the link's length, mass per unit length, Young modulus, and second moment of area are represented by  $L$ ,  $\rho$ ,  $E$ , and  $I$ , respectively. The input torque of joint is  $\tau$  and the lateral displacement of link is defined as  $w(x, t)$ . The mass and the moment of inertia of the endpoint

payload are represented by  $m_e$  and  $J_e$ , respectively. A simplified model of the flexible single-link manipulator is shown in Figure 1.

If the manipulator is considered as a cantilever beam structure and the first two modal shapes of the link are selected, then by the assumed modes method, the lateral displacement  $w(x, t)$  can be expressed as follows:

$$w(x, t) = \phi_1(x)\eta_1(t) + \phi_2(x)\eta_2(t), \quad (1)$$

where the two mode shape functions are selected as follows:

$$\phi_i(x) = ch(\beta_i x) - \cos(\beta_i x) - \frac{ch(\beta_i L) + \cos(\beta_i L)}{sh(\beta_i L) + \sin(\beta_i L)} \left( sh(\beta_i x) - \sin(\beta_i x) \right), \quad i = 1, 2. \quad (2)$$

For any point on the link, the position vector  $\mathbf{r}$  and corresponding velocity vectors  $\dot{\mathbf{r}}$  in inertial coordinate system can be represented, respectively, by:

$$\mathbf{r} = \begin{bmatrix} x \cos \theta - w \sin \theta \\ x \sin \theta + w \cos \theta \end{bmatrix}, \quad (3)$$

$$\dot{\mathbf{r}} = \begin{bmatrix} -\dot{\theta}x \sin \theta - \dot{w} \sin \theta - \dot{\theta}w \cos \theta \\ \dot{\theta}x \cos \theta + \dot{w} \cos \theta - \dot{\theta}w \sin \theta \end{bmatrix}.$$

The system generalized coordinate  $\mathbf{q}$  is selected as  $\mathbf{q} = [\theta \eta_1 \eta_2]^T$ , and the kinetic energy  $T_L$  of the link can be expressed as follows:

$$T_L = \frac{1}{2} \rho \int_0^L \dot{\mathbf{r}}^T \dot{\mathbf{r}} dx$$

$$= \frac{1}{2} \dot{\mathbf{q}}^T \begin{bmatrix} \frac{\rho L^3}{3} + \rho L \eta_1^2 + \rho L \eta_2^2 & 0.5688 \rho L^2 & 0.0908 \rho L^2 \\ 0.5688 \rho L^2 & \rho L & 0 \\ 0.0908 \rho L^2 & 0 & \rho L \end{bmatrix} \dot{\mathbf{q}}$$

$$= \frac{1}{2} \dot{\mathbf{q}}^T \mathbf{M}_1 \dot{\mathbf{q}}, \quad (4)$$

and the kinetic energy  $T_e$  of endpoint payload can be denoted as follows:

$$T_e = \frac{1}{2} m_e \dot{\mathbf{r}}_e^T \dot{\mathbf{r}}_e + \frac{1}{2} J_e (\dot{\theta} + w'(L, t))^2 = \frac{1}{2} \dot{\mathbf{q}}^T \begin{bmatrix} m_e L^2 + m_e (2\eta_1 - 2\eta_2)^2 + J_e & 2m_e L + 2.7528 J_e & -4m_e L - 9.5614 J_e \\ 2m_e L + 2.7528 J_e & 4m_e + 7.5779 J_e & -4m_e - 26.3206 J_e \\ -2m_e L - 9.5614 J_e & -4m_e - 26.3206 J_e & 4m_e + 91.4204 J_e \end{bmatrix} \dot{\mathbf{q}} = \frac{1}{2} \dot{\mathbf{q}}^T \mathbf{M}_2 \dot{\mathbf{q}}, \quad (5)$$

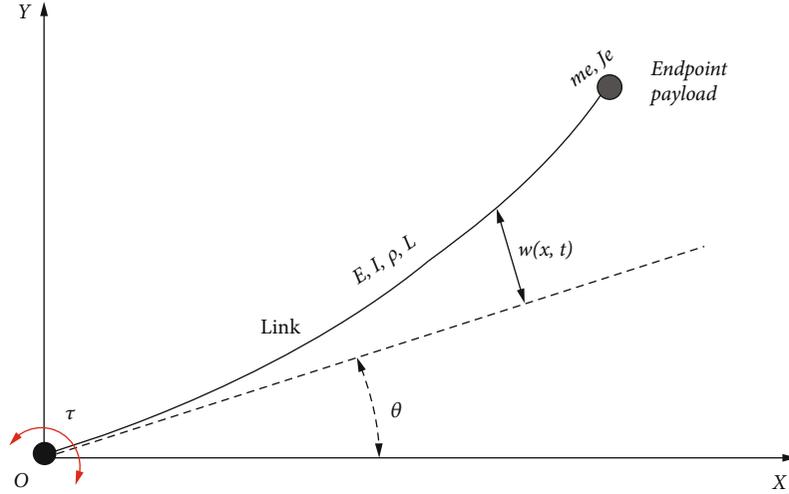


FIGURE 1: Single-link manipulator with endpoint payload.

where  $\mathbf{r}_e$  is the end position vector of the link. Therefore, the system total kinetic energy  $T$  can be expressed as follows:

$$T = T_L + T_e = \frac{1}{2} \dot{\mathbf{q}}^T (\mathbf{M}_1 + \mathbf{M}_2) \dot{\mathbf{q}} = \frac{1}{2} \dot{\mathbf{q}}^T \mathbf{M} \dot{\mathbf{q}}, \quad (6)$$

where  $\mathbf{M}$  is system generalized mass matrix.

The system total potential energy  $U$  only includes the elastic potential energy for the link and thus it can be written as follows:

$$\begin{aligned} U &= \frac{1}{2} EI \int_0^L \left( \frac{\partial^2 w}{\partial x^2} \right)^2 dx \\ &= \frac{1}{2} \mathbf{q}^T \left( EI \int_0^L \begin{bmatrix} 0 & 0 & 0 \\ 0 & (\phi_1'')^2 & \phi_1'' \phi_2'' \\ 0 & \phi_1'' \phi_2'' & (\phi_2'')^2 \end{bmatrix} dx \right) \mathbf{q} = \frac{1}{2} \mathbf{q}^T \mathbf{K} \mathbf{q}, \end{aligned} \quad (7)$$

where  $\mathbf{K}$  is system generalized stiffness matrix. Consequently, based on the Lagrange method, the dynamic equation of the manipulator can be denoted as follows:

$$\mathbf{M} \ddot{\mathbf{q}} + \dot{\mathbf{M}} \dot{\mathbf{q}} - \frac{\partial T}{\partial \mathbf{q}} + \mathbf{K} \mathbf{q} = \mathbf{u}, \quad (8)$$

where  $\dot{\mathbf{M}} \dot{\mathbf{q}}$  and  $\partial T / \partial \mathbf{q}$  are the Coriolis and Centripetal forces, respectively.  $\mathbf{u}$  is the generalized force and  $\mathbf{u} = [\tau \ 0 \ 0]^T$ .

Substituting the generalized mass matrix and stiffness matrix into Equation (8), the system dynamic equation can be rewritten as follows:

$$\mathbf{M}(\mathbf{q}) \ddot{\mathbf{q}} + \mathbf{E}(\mathbf{q}, \dot{\mathbf{q}}) \dot{\mathbf{q}} + \mathbf{K} \mathbf{q} = \mathbf{u}, \quad (9)$$

and the detailed elements in  $n \times n$  matrices  $\mathbf{M}(\mathbf{q})$ ,  $\mathbf{E}(\mathbf{q}, \dot{\mathbf{q}})$ , and  $\mathbf{K}$  are given in Appendix A.

**2.2. Linearization of Nonlinear Dynamic Model.** This paper mainly focuses on the structural vibration and corresponding payload parameter identification problem, so the global motion of the manipulator in the inertial coordinate frame  $O$ - $XYZ$  will be ignored. In this case, based on the local linearization theory, the nonlinear dynamics Equation (9) can be linearized at the selected working point as follows [29]:

$$\mathbf{M}|_{(\mathbf{q})_0} \delta \ddot{\mathbf{q}}(t) + \mathbf{C}_E \delta \dot{\mathbf{q}}(t) + (\mathbf{K} + \mathbf{K}_M + \mathbf{K}_E) \delta \mathbf{q}(t) = \mathbf{L} \mathbf{u}(t), \quad (10)$$

where the notation “ $(\cdot)_0$ ” denotes the value at the working point. Therefore, the linearization state vector  $\delta \mathbf{q}(t) = \mathbf{q}(t) - (\mathbf{q})_0$ ,  $\mathbf{L}$  is the input influence matrix and the input signal  $\mathbf{u}(t) = \tau(t)$ . The linearized matrices  $\mathbf{C}_E$ ,  $\mathbf{K}_M$ , and  $\mathbf{K}_E$  in the equation can be computed, respectively, by:

$$\begin{aligned} \mathbf{C}_E &= \frac{\partial \mathbf{E}}{\partial \dot{\mathbf{q}}}|_{(\mathbf{q})_0, (\dot{\mathbf{q}})_0} = \left[ \frac{\partial \mathbf{E}}{\partial \theta}, \frac{\partial \mathbf{E}}{\partial \eta_1}, \frac{\partial \mathbf{E}}{\partial \eta_2} \right] \Big|_{(\mathbf{q})_0, (\dot{\mathbf{q}})_0}, \\ \mathbf{K}_M &= \frac{\partial \mathbf{M}}{\partial \dot{\mathbf{q}}}|_{(\mathbf{q})_0} (\ddot{\mathbf{q}})_0 = \left[ \frac{\partial \mathbf{M}}{\partial \theta} (\ddot{\mathbf{q}})_0, \frac{\partial \mathbf{M}}{\partial \eta_1} (\ddot{\mathbf{q}})_0, \frac{\partial \mathbf{M}}{\partial \eta_2} (\ddot{\mathbf{q}})_0 \right] \Big|_{(\mathbf{q})_0}, \\ \mathbf{K}_E &= \frac{\partial \mathbf{E}}{\partial \mathbf{q}}|_{(\mathbf{q})_0, (\dot{\mathbf{q}})_0} = \left[ \frac{\partial \mathbf{E}}{\partial \theta}, \frac{\partial \mathbf{E}}{\partial \eta_1}, \frac{\partial \mathbf{E}}{\partial \eta_2} \right] \Big|_{(\mathbf{q})_0, (\dot{\mathbf{q}})_0}. \end{aligned} \quad (11)$$

To simplify the equation derivation, here it is assumed that the modal displacement, velocity, and acceleration of the link at the working point are very small, that is,  $(\eta_i)_0 = (\dot{\eta}_i)_0 = (\ddot{\eta}_i)_0 \approx 0$  and  $(i = 1, 2)$ , so the terms  $\mathbf{C}_E$ ,  $\mathbf{K}_M$ , and  $\mathbf{K}_E$  in Equation (10) will be zero. Then, the linearized dynamics

Equation (10) can be further simplified and the corresponding measurement equation can be expressed, respectively, as follows:

$$\mathbf{M}|_{(\mathbf{q})_0} \delta \ddot{\mathbf{q}}(t) + \mathbf{K} \delta \mathbf{q}(t) = \mathbf{L} \mathbf{u}(t), \quad (12)$$

where

$$\mathbf{M}|_{(\mathbf{q})_0} = \begin{bmatrix} \frac{\rho L^3}{3} + m_e L^2 + J_e & 0.5688 \rho L^2 + 2m_e L + 2.7528 J_e & 0.0908 \rho L^2 - 2m_e L - 9.5614 J_e \\ 0.5688 \rho L^2 + 2m_e L + 2.7528 J_e & \rho L + 4m_e + 7.5779 J_e & -4m_e - 26.3206 J_e \\ 0.0908 \rho L^2 - 2m_e L - 9.5614 J_e & -4m_e - 26.3206 J_e & \rho L + 4m_e + 91.4204 J_e \end{bmatrix}, \quad (14)$$

and  $\mathbf{y}(t)$  is  $m \times 1$  output vector, and  $\mathbf{C}_d$ ,  $\mathbf{C}_v$ , and  $\mathbf{C}_a$  are the  $m \times n$  output influence matrices for displacement, velocity, and acceleration, respectively.

Now, the linearization dynamic equation of the manipulator with payload at the selected working point is obtained. In the following section, the identification of the payload parameters  $\{m_e, J_e\}$  in Equation (12) will be studied by using the system modal matrix estimation. It is worth noting that the linearization Equation (12) can only apply to deal with the small vibration problem at the working point, and it does not suitable to use for the trajectory tracking problem when the manipulator is moving in a wide range.

### 3. Payload Parameter Identification Based on the Modal Matrix Estimation

In this section, Equations (12) and (13) are transformed into the state-space form, and the system state-space model and corresponding eigenvalue matrix are identified by the OKID algorithm. Then, the system mass, stiffness, and damping matrices are computed on the basis of the identified eigenvalues. Finally, the inertial parameters of the endpoint payload in the linearized Equation (12) are obtained using the LS technology.

**3.1. OKID Algorithm.** Define a new state vector  $\mathbf{x}(t) = [\delta \boldsymbol{\eta}^T(t) \quad \delta \dot{\boldsymbol{\eta}}^T(t)]^T$  and then Equations (12) and (13) can be rewritten as the following state-space form:

$$\dot{\mathbf{x}}(t) = \mathbf{A}_c \mathbf{x}(t) + \mathbf{B}_c \mathbf{u}(t), \quad (15)$$

$$\mathbf{y}(t) = \mathbf{C} \mathbf{x}(t) + \mathbf{D} \mathbf{u}(t), \quad (16)$$

where  $\mathbf{A}_c$ ,  $\mathbf{B}_c$ ,  $\mathbf{C}$ , and  $\mathbf{D}$  are the  $2n \times 2n$  system,  $2n \times r$  input,  $m \times 2n$  output, and  $m \times r$  direct transmission matrices in the continuous system, respectively:

$$\mathbf{y}(t) = \mathbf{C}_a \delta \ddot{\mathbf{q}}(t) + \mathbf{C}_v \delta \dot{\mathbf{q}}(t) + \mathbf{C}_d \delta \mathbf{q}(t), \quad (13)$$

$$\begin{aligned} \mathbf{A}_c &= \begin{bmatrix} 0 & \mathbf{I} \\ -\mathbf{M}|_{(\mathbf{q})_0}^{-1} \mathbf{K} & 0 \end{bmatrix}, \\ \mathbf{B}_c &= \begin{bmatrix} 0 \\ \mathbf{M}|_{(\mathbf{q})_0}^{-1} \mathbf{L} \end{bmatrix}, \\ \mathbf{C} &= [\mathbf{C}_d - \mathbf{C}_a \mathbf{M}^{-1} \mathbf{K} \quad 0], \\ \mathbf{D} &= \mathbf{C}_a \mathbf{M}^{-1} \mathbf{L}, \end{aligned} \quad (17)$$

where  $\mathbf{I}$  is the unit matrix. Since the following computations are implemented under discrete systems, Equations (15) and (16) of the continuous system can be further transformed into the discrete state-space equations of the following form as:

$$\mathbf{x}(k+1) = \mathbf{A} \mathbf{x}(k) + \mathbf{B} \mathbf{u}(k), \quad (18)$$

$$\mathbf{y}(k) = \mathbf{C} \mathbf{x}(k) + \mathbf{D} \mathbf{u}(k), \quad (19)$$

where  $\mathbf{A}$  and  $\mathbf{B}$  are the discretized system and input matrices, respectively. Further, the observer form of Equation (18) can be expressed as follows:

$$\mathbf{x}(k+1) = \bar{\mathbf{A}} \mathbf{x}(k) + \bar{\mathbf{B}} \mathbf{v}(k), \quad (20)$$

where

$$\mathbf{v}(k) = \begin{bmatrix} \mathbf{u}(k) \\ \mathbf{y}(k) \end{bmatrix}, \quad (21)$$

where  $\bar{\mathbf{G}}$  is an  $2n \times m$  arbitrary matrix chosen to make the matrix  $\bar{\mathbf{A}}$  as stable as desired. Then, for the time step  $k+1$ ,  $k+2$ ,  $\dots$ ,  $k+p$ , the measurement Equation (19) can be expressed, respectively, as follows:

$$\begin{aligned}
\mathbf{y}(k+1) &= \mathbf{C}\bar{\mathbf{A}}\mathbf{x}(k) + \mathbf{C}\bar{\mathbf{B}}\mathbf{v}(k) + \mathbf{D}\mathbf{u}(k+1), \\
\mathbf{y}(k+2) &= \mathbf{C}\bar{\mathbf{A}}^2\mathbf{x}(k) + \mathbf{C}\bar{\mathbf{A}}\bar{\mathbf{B}}\mathbf{v}(k) + \bar{\mathbf{B}}\mathbf{v}(k+1) + \mathbf{D}\mathbf{u}(k+2), \\
&\vdots \\
\mathbf{y}(k+p) &= \mathbf{C}\bar{\mathbf{A}}^p\mathbf{x}(k) + \mathbf{C}\bar{\mathbf{A}}^{p-1}\bar{\mathbf{B}}\mathbf{v}(k) + \dots + \mathbf{C}\bar{\mathbf{B}}\mathbf{v}(k+p-1) + \mathbf{D}\mathbf{u}(k+p),
\end{aligned} \tag{22}$$

and the set of these equations for a sequence of  $k = 0, 1, \dots, l-1$  can be written as follows:

$$\bar{\mathbf{y}} = \mathbf{C}\bar{\mathbf{A}}^p\mathbf{x} + \bar{\mathbf{Y}}\bar{\mathbf{V}}, \tag{23}$$

where

$$\begin{aligned}
\bar{\mathbf{y}} &= [\mathbf{y}(p), \mathbf{y}(p+1), \dots, \mathbf{y}(l-1)], \\
\bar{\mathbf{Y}} &= [\mathbf{D}, \mathbf{C}\bar{\mathbf{B}}, \mathbf{C}\bar{\mathbf{A}}\bar{\mathbf{B}}, \dots, \mathbf{C}\bar{\mathbf{A}}^{p-1}\bar{\mathbf{B}}], \\
\bar{\mathbf{V}} &= \begin{bmatrix} \mathbf{u}(p) & \mathbf{u}(p+1) & \dots & \mathbf{u}(l-1) \\ \mathbf{v}(p-1) & \mathbf{v}(p) & \dots & \mathbf{v}(l-2) \\ \vdots & \vdots & \ddots & \vdots \\ \mathbf{v}(0) & \mathbf{v}(1) & \dots & \mathbf{v}(l-p-1) \end{bmatrix},
\end{aligned} \tag{24}$$

where  $\mathbf{D}$ ,  $\mathbf{C}\bar{\mathbf{B}}$ ,  $\mathbf{C}\bar{\mathbf{A}}\bar{\mathbf{B}}$ ,  $\mathbf{L}$ ,  $\mathbf{C}\bar{\mathbf{A}}^{p-1}\bar{\mathbf{B}}$  are called observer Markov parameters. The first term in Equation (23) represents the effect of the preceding  $p-1$  time steps. It can select the proper matrix  $\mathbf{G}$  to ensure that the observer model  $\bar{\mathbf{A}}^p$  is asymptotically stable and all the states in the vector  $\mathbf{x}$  are bounded, so if the time step is long enough,  $\bar{\mathbf{A}}^p \approx 0$  and thus Equation (23) can be approximated by neglecting the first term on the right-hand side as follows:

$$\bar{\mathbf{y}}_{m \times l} = \bar{\mathbf{Y}}_{m \times [p(m+r)+r]} \bar{\mathbf{V}}_{[p(m+r)+r] \times l}. \tag{25}$$

Then, the least-squares solution of the observer Markov parameter matrix  $\bar{\mathbf{Y}}$  in Equation (25) is as follows:

$$\bar{\mathbf{Y}} = \bar{\mathbf{y}}\bar{\mathbf{V}}^\dagger, \tag{26}$$

where the notation “ $\dagger$ ” is the Moore-Penrose inverse.

If define the observer Markov parameters  $\mathbf{D}$ ,  $\mathbf{C}\bar{\mathbf{B}}$ ,  $\mathbf{C}\bar{\mathbf{A}}\bar{\mathbf{B}}$ ,  $\dots$ ,  $\mathbf{C}\bar{\mathbf{A}}^{p-1}\bar{\mathbf{B}}$  in matrix  $\bar{\mathbf{Y}}$  as the following form:

$$\bar{\mathbf{Y}} = [\mathbf{D}, \mathbf{C}\bar{\mathbf{B}}, \mathbf{C}\bar{\mathbf{A}}\bar{\mathbf{B}}, \dots, \mathbf{C}\bar{\mathbf{A}}^{p-1}\bar{\mathbf{B}}] = [\bar{\mathbf{Y}}_0, \bar{\mathbf{Y}}_1, \bar{\mathbf{Y}}_2, \dots, \bar{\mathbf{Y}}_p], \tag{27}$$

then the observer Markov parameters  $\bar{\mathbf{Y}}_i$  for each time step  $k = 0, 1, 2, \dots$  can be computed by:

$$\begin{aligned}
\bar{\mathbf{Y}}_0 &= \mathbf{D}, \\
\bar{\mathbf{Y}}_k &= \mathbf{C}\bar{\mathbf{A}}^{k-1}\bar{\mathbf{B}} = [\mathbf{C}(\mathbf{A} + \mathbf{G}\mathbf{C})^{k-1}(\mathbf{B} + \mathbf{G}\mathbf{D}), -\mathbf{C}(\mathbf{A} + \mathbf{G}\mathbf{C})^{k-1}\mathbf{G}] \\
&\triangleq [\bar{\mathbf{Y}}_k^{(1)}, -\bar{\mathbf{Y}}_k^{(2)}], \quad k = 1, 2, \dots
\end{aligned} \tag{28}$$

Based on the definition of Markov parameters, the system Markov parameters  $\mathbf{Y}_i (i = 1, 2, \dots)$  can be recovered from the observer Markov parameters  $\bar{\mathbf{Y}}_i$  by using the following relationship:

$$\begin{aligned}
\mathbf{Y}_0 &= \bar{\mathbf{Y}}_0 = \mathbf{D}, \\
\mathbf{Y}_k &= \mathbf{Y}_k^{(1)} - \sum_{i=1}^k \bar{\mathbf{Y}}_i^{(2)} \mathbf{Y}_{k-i}, \quad \text{for } k = 1, 2, \dots, p, \\
\mathbf{Y}_k &= -\sum_{i=1}^p \bar{\mathbf{Y}}_i^{(2)} \mathbf{Y}_{k-i}, \quad \text{for } k = p+1, p+2, \dots, \infty.
\end{aligned} \tag{29}$$

Using the system Markov parameters  $\mathbf{Y}_k$ , the system Hankel matrix  $\mathbf{H}$  can be written as follows:

$$\mathbf{H}(k-1) = \begin{bmatrix} \mathbf{Y}_k & \mathbf{Y}_{k+1} & \dots & \mathbf{Y}_{k+\beta-1} \\ \mathbf{Y}_{k+1} & \mathbf{Y}_{k+2} & \dots & \mathbf{Y}_{k+\beta} \\ \vdots & \vdots & \ddots & \vdots \\ \mathbf{Y}_{k+\alpha-1} & \mathbf{Y}_{k+\alpha} & \dots & \mathbf{Y}_{k+\alpha+\beta-2} \end{bmatrix}_{m \times r \beta}, \tag{30}$$

where  $\alpha$  and  $\beta$  are arbitrary positive integers, and the rank of the matrix  $\mathbf{H}(k-1)$  should be larger than the system order  $2n$  of the state-space model [22].

Then, the SVD is implemented for the Hankel matrix  $\mathbf{H}(0)$ :

$$\mathbf{H}(0) = \mathbf{R}\mathbf{\Sigma}\mathbf{S}^T, \tag{31}$$

where the columns of the matrices  $\mathbf{R}$  and  $\mathbf{S}$  are orthonormal, and  $\mathbf{\Sigma}$  is a rectangular matrix as follows:

$$\mathbf{\Sigma} = \begin{bmatrix} \sum_{2n} & 0 \\ 0 & 0 \end{bmatrix}, \tag{32}$$

with  $\sum_{2n}$  is a diagonal matrix which includes the system singular values  $\sigma_1, \sigma_2, \dots, \sigma_{2n}$ .

Therefore, the desired minimum realization matrices  $\{\mathbf{A}, \mathbf{B}, \mathbf{C}, \mathbf{D}\}$  of this state-space system can be obtained by [30]:

$$\hat{\mathbf{A}} = \sum_{2n}^{-1/2} \mathbf{R}_{2n}^T \mathbf{H}(1) \mathbf{S}_{2n} \sum_{2n}^{-1/2}, \tag{33}$$

$$\hat{\mathbf{B}} = \sum_{2n}^{1/2} \mathbf{S}_{2n}^T \mathbf{E}_r, \tag{34}$$

$$\hat{\mathbf{C}} = \mathbf{E}_m^T \mathbf{R}_{2n} \sum_{2n}^{1/2}, \tag{35}$$

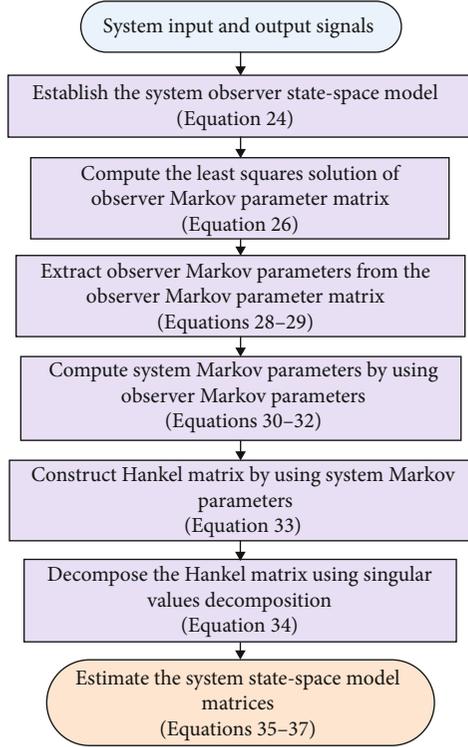


FIGURE 2: Computation procedure of the OKID algorithm.

where the superscript “ $\wedge$ ” in Equations (33)–(35) denotes the identified values of relevant matrices that are distinguished from the original values  $\{\mathbf{A}, \mathbf{B}, \mathbf{C}\}$ , and the matrices  $\mathbf{R}_{2n}$ ,  $\mathbf{S}_{2n}$ ,  $\mathbf{E}_r$ , and  $\mathbf{E}_m$  in Equations (33)–(35) are represented by:

$$\mathbf{R}_{2n} = \mathbf{R}(:, 1 : 2n), \mathbf{S}_{2n} = \mathbf{S}(:, 1 : 2n), \mathbf{E}_r = \begin{bmatrix} \mathbf{I}_{r \times r} & \mathbf{0}_{r \times [r(\beta-1)]} \end{bmatrix}^T, \mathbf{E}_m = \begin{bmatrix} \mathbf{I}_{m \times m} & \mathbf{0}_{m \times [m(\alpha-1)]} \end{bmatrix}^T, \quad (36)$$

and the expression “ $(:, 1 : 2n)$ ” denotes that the first  $2n$  columns of the matrix are selected. The brief computation flow chart of system state-space model by using the OKID algorithm is shown in Figure 2.

**3.2. Similarity Transformation of State-Space Model.** In the previous section, the OKID algorithm was used to obtain a set of identified discrete model parameters  $\{\hat{\mathbf{A}}, \hat{\mathbf{B}}, \hat{\mathbf{C}}\}$ . The two sets of state-space models, namely,  $\{\hat{\mathbf{A}}, \hat{\mathbf{B}}, \hat{\mathbf{C}}\}$  and  $\{\mathbf{A}, \mathbf{B}, \mathbf{C}\}$ , have the same system I–O relationship, but the detailed element values in the two sets of models, such as the values of matrices  $\hat{\mathbf{A}}$  and  $\mathbf{A}$ , are different. On this basis, a system includes an infinite set of the state-space model  $\{\mathbf{A}_i, \mathbf{B}_i, \mathbf{C}_i\}$ ,  $(i = 1, 2, \dots, \infty)$ . Therefore, when a set of the state-space model is identified by the OKID algorithm, if the detailed elements for the original state-space model are estimated, then a similar transformation should be first performed between the original and identified state-space models by using a transformation matrix to obtain the original model parameters  $\{\mathbf{A}, \mathbf{B}, \mathbf{C}\}$  [31]. In this case,

the identified and original state-space models have the following relationships:

$$\begin{cases} \hat{\mathbf{A}} = \mathbf{T}\mathbf{A}\mathbf{T}^{-1}, \\ \hat{\mathbf{B}} = \mathbf{T}\mathbf{B}, \\ \hat{\mathbf{C}} = \mathbf{C}\mathbf{T}^{-1}, \end{cases} \quad (37)$$

where  $\mathbf{T}$  is the transformation matrix. In practical situation, the original input matrix  $\mathbf{C}$  is usually priori known. For example, Equation (17) indicates that the matrix  $\mathbf{C}$  was constructed by the output influence matrices  $\mathbf{C}_d$ ,  $\mathbf{C}_v$ , and  $\mathbf{C}_a$ , which were obtained from certain prior knowledge, such as finite element analysis. Therefore,  $\mathbf{T} = \mathbf{C}\mathbf{C}^\dagger$ , and the original state-space model  $\{\mathbf{A}, \mathbf{B}, \mathbf{C}\}$  was determined by Equation (37).

**3.3. Identification of System Modal Parameters.** The system state-space model Equations (18) and (19) can be decoupled into  $n$  pairs of complex conjugate modes by solving the eigenproblem of  $\mathbf{A}$ . Therefore, the system’s modal parameters (frequencies, damping ratios, and mode shape matrices) can be computed from the system state-space model matrices  $\{\mathbf{A}, \mathbf{C}\}$ . Firstly, the eigenvalue decomposition of system matrix  $\mathbf{A}$  is as follows [32]:

$$\mathbf{A} = \mathbf{\Psi} \mathbf{\Sigma} \mathbf{\Psi}^{-1}, \quad (38)$$

where

$$\mathbf{\Sigma} = \begin{bmatrix} \Lambda_d & \\ & \Lambda_d^* \end{bmatrix} \in \mathbb{R}^{2n \times 2n} \quad (39)$$

is a diagonal matrix with complex eigenvalue pairs  $\{\lambda_j, \lambda_j^*\}$ ,  $(j = 1, 2, \dots, n)$ , where the notation “ $*$ ” denotes the conjugate.  $\mathbf{\Psi} \in \mathbb{R}^{2n \times 2n}$  is the corresponding complex eigenvector matrix.

Then, the complex eigenvalues  $\lambda_{cj}$ ,  $(j = 1, 2, \dots, n)$  for the continuous-time state-space model (15) and (16) can be expressed as follows:

$$\lambda_{cj} = \frac{1}{\Delta t} \ln \lambda_j, \quad (40)$$

where  $\Delta t$  is the sampling time. Consequently, the  $j$ th normal undamped circular frequency  $\omega_j$ , damping ratio  $\xi_j$ , and experimental mode shapes matrix  $\Phi$  at the sensor locations for the dynamic system (12) and (13) can be obtained, respectively, as follows:

$$\begin{aligned} \omega_j &= \sqrt{(\operatorname{Re}(\lambda_{cj}))^2 + (\operatorname{Im}(\lambda_{cj}))^2}, \\ \xi_j &= \frac{|\operatorname{Re}(\lambda_{cj})|}{\sqrt{(\operatorname{Re}(\lambda_{cj}))^2 + (\operatorname{Im}(\lambda_{cj}))^2}}, \end{aligned} \quad (41)$$

$$\Phi = \mathbf{C}\mathbf{\Psi},$$

where  $\operatorname{Re}(\lambda_{cj})$  and  $\operatorname{Im}(\lambda_{cj})$  denote the real part and imaginary part of the complex eigenvalue  $\lambda_{cj}$ , respectively.

Therefore, for the continuous-time system model (15) and (16), the complex eigenvalue matrix  $\Lambda_c \in \mathbb{R}^{n \times n}$  and corresponding complex eigenvector matrix  $\Psi_c$  can be expressed as follows:

$$\begin{aligned}\Lambda_c &= \text{diag} \{ \lambda_{cj} \}, \\ \Lambda_c^* &= \text{diag} \{ \lambda_{cj}^* \}, \\ \Psi_c &= \Psi = \begin{bmatrix} \Psi & \Psi^* \\ \Psi \Lambda_c & \Psi^* \Lambda_c^* \end{bmatrix},\end{aligned}\quad (42)$$

where  $\Psi_c$  is a  $2n \times 2n$  matrix containing the corresponding complex mode shape  $\Psi \in \mathbb{R}^{n \times n}$ , which can be expressed as follows:

$$\Psi = [\Psi_1 \quad \Psi_2 \quad \cdots \quad \Psi_n] = \begin{bmatrix} \Psi_{11} & \Psi_{12} & \cdots & \Psi_{1n} \\ \Psi_{21} & \Psi_{22} & \cdots & \Psi_{2n} \\ \vdots & \cdots & \ddots & \vdots \\ \Psi_{n1} & \Psi_{n2} & \cdots & \Psi_{nn} \end{bmatrix}.\quad (43)$$

Then, the complex eigenvalue matrix  $\Lambda_c$  and complex eigenvector  $\Psi_c$  for the dynamics Equations (12) and (13) are determined by using the OKID algorithm. In the next section, the parameters  $\{\Lambda_c, \Psi_c\}$  will be used to estimate the inertia parameters  $m_e$  and  $J_e$  of the endpoint payload object based on complex modal analysis [33, 34].

**3.4. Inertia Parameter Estimation of Endpoint Payload Object.** After the identification of complex eigenvalue  $\Lambda_c$  and corresponding eigenvector matrices  $\Psi_c$  of the dynamic system (12) and (13), Equation (12) can be written using a canonical form by

$$\begin{bmatrix} 0 & \mathbf{M}|_{(q)_0} \\ \mathbf{M}|_{(q)_0} & 0 \end{bmatrix} \begin{bmatrix} \delta \dot{\mathbf{q}}(t) \\ \delta \ddot{\mathbf{q}}(t) \end{bmatrix} = \begin{bmatrix} -\mathbf{K} & 0 \\ 0 & \mathbf{M}|_{(q)_0} \end{bmatrix} \begin{bmatrix} \delta \mathbf{q}(t) \\ \delta \dot{\mathbf{q}}(t) \end{bmatrix} + \begin{bmatrix} \mathbf{L} \\ 0 \end{bmatrix} \mathbf{u}(t).\quad (44)$$

Furthermore, the generalized eigenproblem can be written in a symmetric form as follows:

$$\begin{bmatrix} -\mathbf{K} & 0 \\ 0 & \mathbf{M}|_{(q)_0} \end{bmatrix} \begin{bmatrix} \Psi \\ \Psi \Lambda_c \end{bmatrix} = \begin{bmatrix} 0 & \mathbf{M}|_{(q)_0} \\ \mathbf{M}|_{(q)_0} & 0 \end{bmatrix} \begin{bmatrix} \Psi \\ \Psi \Lambda_c \end{bmatrix} \Lambda_c.\quad (45)$$

By simple algebraic manipulations, the orthogonality conditions in Equation (45) can be represented as the following equivalent eigenvalue expression:

$$\mathbf{M}|_{(q)_0} \Psi \Lambda_c^2 + \mathbf{K} \Psi \Lambda_c = 0.\quad (46)$$

The symmetric formulation in Equation (46) leads to a general solution for the inverse damped vibration problem. Then, the following  $n$  complex eigenequations can be obtained:

$$\left( \lambda_{cj}^2 \mathbf{M}|_{(q)_0} + \mathbf{K} \right) \Psi_j = 0_{n \times 1}, \quad (j = 1, 2, \dots, n).\quad (47)$$

For an  $n$ -order system as Equation (47), if the generalized form of the elements in the matrices  $\mathbf{M}|_{(q)_0}$  and  $\mathbf{K}$  is defined as follows, then

$$\begin{aligned}\mathbf{M}|_{(q)_0} &= \begin{bmatrix} m_{11} & m_{12} & \cdots & m_{1n} \\ m_{21} & m_{22} & \cdots & m_{2n} \\ \vdots & \vdots & \ddots & \vdots \\ m_{n1} & m_{n2} & \cdots & m_{nn} \end{bmatrix}, \\ \mathbf{K} &= \begin{bmatrix} k_{11} & & & \\ & k_{22} & & \\ & & \ddots & \\ & & & k_{nn} \end{bmatrix},\end{aligned}\quad (48)$$

and the complex mode shape  $\Psi_j \in \mathbb{R}^{n \times 1}$ , ( $j = 1, 2, \dots, n$ ) in Equation (42) can be expressed as follows:

$$\Psi_j = \begin{bmatrix} \Psi_{1j} \\ \Psi_{2j} \\ \vdots \\ \Psi_{nj} \end{bmatrix}, \quad (j = 1, 2, \dots, n).\quad (49)$$

Therefore, the corresponding generalized form of Equation (47) can be further expressed as follows:

$$\begin{aligned}& \left( \lambda_{cj}^2 \mathbf{M}|_{(q)_0} + \mathbf{K} \right) \Psi_j \\ &= \left( \lambda_{cj}^2 \begin{bmatrix} m_{11} & m_{12} & \cdots & m_{1n} \\ m_{21} & m_{22} & \cdots & m_{2n} \\ \vdots & \vdots & \ddots & \vdots \\ m_{n1} & m_{n2} & \cdots & m_{nn} \end{bmatrix} + \begin{bmatrix} k_{11} & & & \\ & k_{22} & & \\ & & \ddots & \\ & & & k_{nn} \end{bmatrix} \right) \begin{bmatrix} \Psi_{1j} \\ \Psi_{2j} \\ \vdots \\ \Psi_{nj} \end{bmatrix}\end{aligned}$$

$$\begin{aligned}
&= \begin{bmatrix} \lambda_{c_j}^2 m_{11} \psi_{1j} + \lambda_{c_j}^2 m_{12} \psi_{2j} + \dots + \lambda_{c_j}^2 m_{1n} \psi_{nj} + k_{11} \psi_{1j} \\ \lambda_{c_j}^2 m_{21} \psi_{1j} + \lambda_{c_j}^2 m_{22} \psi_{2j} + \dots + \lambda_{c_j}^2 m_{2n} \psi_{nj} + k_{22} \psi_{2j} \\ \vdots \\ \lambda_{c_j}^2 m_{n1} \psi_{1j} + \lambda_{c_j}^2 m_{n2} \psi_{2j} + \dots + \lambda_{c_j}^2 m_{nn} \psi_{nj} + k_{nn} \psi_{nj} \end{bmatrix} \\
&= 0_{n \times 1}, \quad (j = 1, 2, \dots, n).
\end{aligned} \tag{50}$$

The present study shows the system order  $n = 3$  in Equation (47), and three eigenvalues, namely,  $\lambda_{c1}$ ,  $\lambda_{c2}$ , and  $\lambda_{c3}$ , are available. Considering that the element  $k_{11}$  in the stiffness matrix  $\mathbf{K}$  is equal to zero (the relevant element values in Appendix A), the first eigenvalue  $\lambda_{c1}$  is zero, and the two other eigenvalues  $\lambda_{c2}$  and  $\lambda_{c3}$  are nonzero. Therefore, based on the description elements of mass matrix  $\mathbf{M}|_{(q_0)}$  in Equation (12), Equation (50) for the arbitrary nonzero eigenvalues (e.g.,  $\lambda_{c2}$ ) can be denoted as follows:

$$\begin{aligned}
&(\lambda_{c2}^2 \mathbf{M}|_{(q_0)} + \mathbf{K}) \psi_2 \\
&= \begin{bmatrix} \lambda_{c2}^2 m_{11} \psi_{12} + \lambda_{c2}^2 m_{12} \psi_{22} + \lambda_{c2}^2 m_{13} \psi_{32} + k_{11} \psi_{12} \\ \lambda_{c2}^2 m_{21} \psi_{12} + \lambda_{c2}^2 m_{22} \psi_{22} + \lambda_{c2}^2 m_{23} \psi_{32} + k_{22} \psi_{22} \\ \lambda_{c2}^2 m_{31} \psi_{12} + \lambda_{c2}^2 m_{32} \psi_{22} + \lambda_{c2}^2 m_{33} \psi_{32} + k_{33} \psi_{32} \end{bmatrix} = 0_{3 \times 1},
\end{aligned} \tag{51}$$

where  $m_{11}, m_{12}, \dots, m_{33}$  can refer to Equation (12) which contains the unknown parameters  $m_e$  and  $J_e$  that must be identified.

In Equation (51), the physical parameters  $\rho$  and  $L$  of the link and the stiffness  $\{k_{11}, k_{22}, k_{33}\}$ , respectively, in matrix  $\mathbf{K}$  are constant. Thus, these parameters can be regarded as known prior knowledge, and the eigenvalue  $\lambda_{c2}$  and complex mode shape  $\psi_2 = [\psi_{12} \ \psi_{22} \ \psi_{32}]^T$  can be identified by the OKID method. Consequently, only two unknown payload parameters,  $m_e$  and  $J_e$ , must be identified. Equation (51) can be further written to the following generalized LS form:

$$\Xi \mathbf{w} = \boldsymbol{\delta}, \tag{52}$$

where

$$\begin{aligned}
\Xi &\triangleq \begin{bmatrix} a_{11} & a_{12} \\ a_{21} & a_{22} \\ a_{31} & a_{32} \end{bmatrix}, \\
\mathbf{w} &= \begin{bmatrix} m_e \\ J_e \end{bmatrix}, \\
\boldsymbol{\delta} &= \begin{bmatrix} b_1 \\ b_2 \\ b_3 \end{bmatrix}.
\end{aligned} \tag{53}$$

The detailed element values in matrix  $\Xi$  and vector  $\boldsymbol{\delta}$  are presented in Appendix B. By using the following LS method,  $m_e$  and  $J_e$  can then be estimated as follows:

$$\hat{\mathbf{w}} = \begin{bmatrix} \hat{m}_e \\ \hat{J}_e \end{bmatrix} = \Xi^\dagger \boldsymbol{\delta}, \tag{54}$$

where the notation “ $\wedge$ ” denotes the identified value. Therefore, on the basis of Equation (54), the identification of the payload parameters  $m_e$  and  $J_e$  is reduced to an LS problem with a unique solution.

In Equation (51), only two parameters  $m_e$  and  $J_e$  must be identified in this study and system order  $n = 3$  indicates that three modal equations can be used to estimate the two unknown parameters. If certain prior knowledge in the mass matrix  $\mathbf{M}|_{(q_0)}$ , such as the length  $L$  of the link and the coefficients of  $m_e$  and  $J_e$ , can be obtained by system modeling, then the identified values  $\hat{m}_e$  and  $\hat{J}_e$  can be theoretically calculated only using arbitrary one system eigenvalue (e.g.,  $\lambda_{c2}$  or  $\lambda_{c3}$ ). However, in practical application, the influence of modeling error and the model order truncation is observed, and then the coefficients of  $m_e$  and  $J_e$  in the mass matrix  $\mathbf{M}|_{(q_0)}$  may not be the same as the theory values. Therefore, all the non-zero eigenvalues and corresponding complex mode shapes of the system can also be used to obtain multiple modal equations, as in Equation (51). Then, the LS estimation in Equation (52) can be implemented. This estimation contains all the system modal information to overcome the inaccuracy problem of the matrix coefficients as much as possible.

**3.5. Summary of Identification Procedures for Payload Parameters.** The parameter identification procedures of the endpoint payload object of the manipulator system can be summarized as follows:

*Step 1.* According to the input and output signals, the system model parameters  $\{\mathbf{A}, \mathbf{B}, \mathbf{C}\}$  are estimated using the OKID method and matrix similarity transformation by Equations (33), (34), (35), and (37).

*Step 2.* The complex eigenvalues  $\lambda_c$  and complex mode shape  $\psi$  of the dynamics Equations (12) and (13) are determined from the identified system model parameters  $\{\mathbf{A}, \mathbf{B}, \mathbf{C}\}$  by Equations (38), (40), (42), and (43).

*Step 3.* Using the complex eigenvalues  $\lambda_c$  and complex mode shape  $\psi$ , the complex eigenequations of the manipulator system operates with payload can be established by Equations (47) and (51).

*Step 4.* The identified mass  $\hat{m}_e$  and moment of inertia  $\hat{J}_e$  of unknown endpoint payload can be computed based on the least-squares estimation of Equation (54).

## 4. Numerical Simulations

In simulations, a flexible single-link manipulator model is first established, and the designed motor torques and the

computed system responses are used as the I–O signals for identification. Subsequently, the system state-space model and corresponding eigenvalue matrix are determined using the OKID algorithm. Finally, the identification of payload parameters is performed using the presented approach, the RLS method, and the APSA algorithm.

**4.1. Simulation Parameters.** The structural parameters of the manipulator are listed in Table 1, where the original values of the payload mass  $m_e$  and moment of inertia  $J_e$  are 100 kg and 200 kg m<sup>2</sup>, respectively. The system working point  $(\mathbf{q})_0 = [(\theta)_0, (\eta_1)_0, (\eta_2)_0]$  is selected as  $(\theta)_0 = 30^\circ$  and  $(\eta_1)_0 = (\eta_2)_0 = 0$ .

The input torque  $\tau$  of the motor is designed as a sinusoidal signal  $\tau = 50 \sin(3.5t)$  ( $0 \leq t \leq 1.8$  s) as shown in Figure 3 to ensure that the manipulator motion is near the working point and the original nonlinear model can be linearized. Identification under disturbance is a disadvantage of the OKID method. The OKID method is an LS method that aims at minimizing the error between real and identified outputs. Therefore, if the SNR is low, then the noise usually has high energy. Moreover, distinguishing the noise from the true system signals using the ERA–OKID series method is difficult. Thus, only 5% of Gaussian random measurement noise is added in the system.

For the working point  $\{(\theta)_0 = 30^\circ, (\eta_1)_0 = (\eta_2)_0 = 0\}$ , the corresponding system responses without and with payload are shown in Figures 4 and 5, respectively. The original nonlinear dynamic equation can be linearized in this I–O condition due to the small system outputs (the response signal of the rotation angle  $\delta\theta$  is less than  $5^\circ$ ).

**4.2. Identification of System State-Space Model and Frequencies.** Using the designed I–O signals in Section 4.1, the state-space model of the manipulator with payload is identified using the OKID algorithm, in which the parameters of the OKID algorithm are  $\alpha = \beta = 5$ , and the sampling time is  $\Delta t = 0.001$  s. Then, the prediction errors of the first three system output signals  $\{y_1, y_2, y_3\}$  using the OKID algorithm are provided in Figure 6(a), and corresponding variance is shown in Figure 6(b). Moreover, the singular value magnitude of the system is shown as Figure 7. In Figure 7, it is easy to determine the system order  $n$  is  $n = 6$  from the truncation of the singular value curve.

According to the similarity transformation mentioned in Section 3.2, the different sets of system state-space models satisfy the same I–O relationship. Therefore, the same test inputs are applied to the original and identified state-space models with the zero initial state condition to verify the accuracy of the identified model. The corresponding system responses for the two sets of state-space models at selected working points are shown in Figure 8. The test responses of the estimated state-space models are consistent with those of the original system. The results verify that the identified system model parameters are reasonably accurate.

Using the identified state-space model, the values of the 2nd and 3rd eigenvalues of the system are computed and presented in Table 2, and the corresponding complex mode shapes are provided in Table 3. The values of 1st-order eigen-

TABLE 1: Physical parameters of the flexible single-link manipulator for simulations.

Component	Symbol	Value	Unit
Link length	$L$	1.00	(m)
Length density of link	$\rho$	19.5	(kg m <sup>-1</sup> )
Elastic modulus of link	$E$	20.00	(GPa)
Inertia moment of link	$I$	$5.20 \times 10^{-7}$	(m <sup>4</sup> )
Payload mass	$m_e$	100.00	(kg)
Payload moment of inertia	$J_e$	200.00	(kg m <sup>2</sup> )

value and mode shapes are zero because they denote the rigid rotation motion of the manipulator. The results in Tables 2 and 3, respectively, illustrate that the algorithm can effectively identify the eigenvalues and mode shapes of the manipulator system at the working point.

**4.3. Identification of Payload Parameters.** The proposed method in Section 3.4 is implemented once the state-space model and complex eigenvalue parameters are obtained using the OKID algorithm to determine the inertia parameters  $\{m_e, J_e\}$  of payload. In addition to the proposed algorithm, the RLS and APSA methods are also employed in this simulation to identify the payload parameters.

For Equation (52), the standard regression form of the RLS algorithm is as follows:

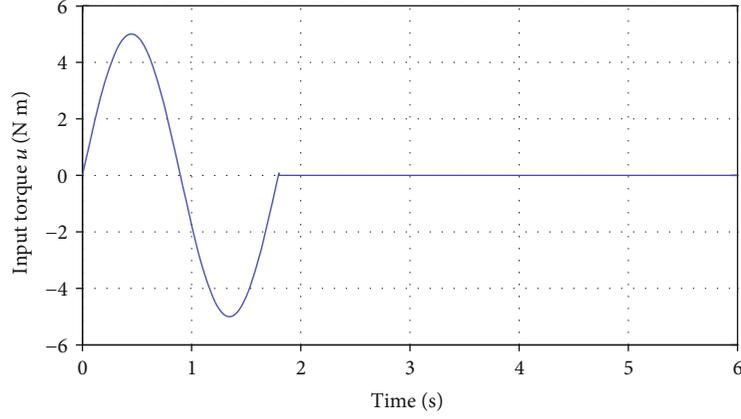
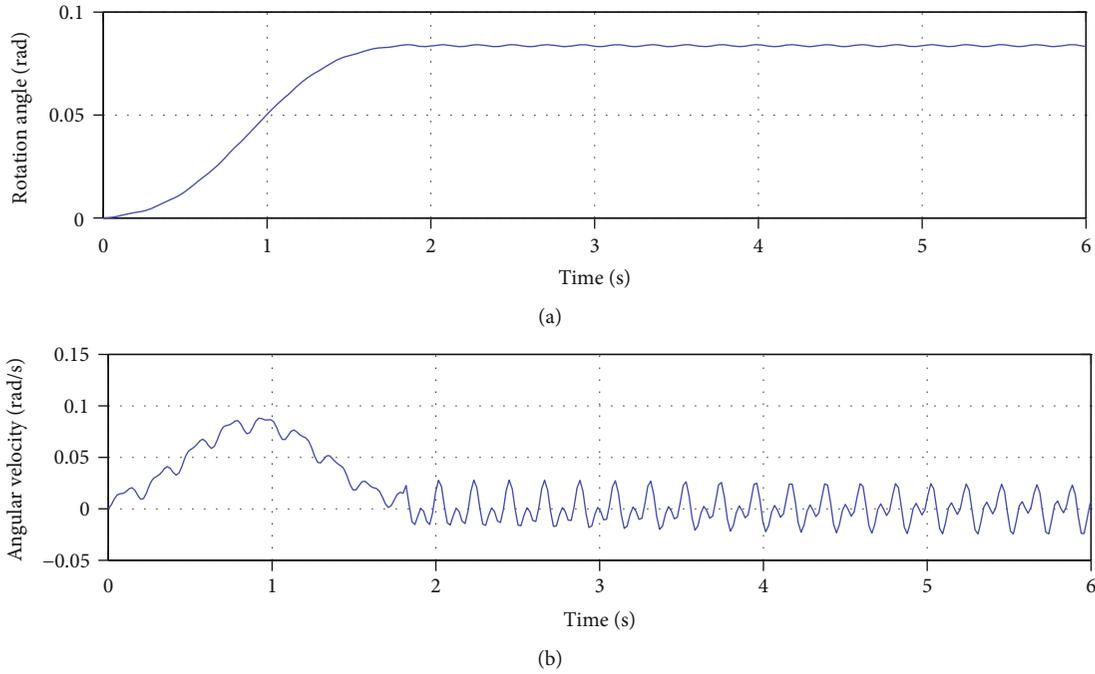
$$\begin{aligned}\widehat{\mathbf{w}}(t+1) &= \widehat{\mathbf{w}}(t) + \mathbf{W}(t)[\boldsymbol{\delta}(t+1) - \boldsymbol{\Xi}(t+1)\widehat{\mathbf{w}}(t)], \\ \mathbf{W}(t+1) &= \mathbf{Z}(t)\boldsymbol{\Xi}^T(t+1)[\gamma\mathbf{I} + \boldsymbol{\Xi}(t+1)\mathbf{Z}(t)\boldsymbol{\Xi}^T(t+1)]^{-1}, \\ \mathbf{Z}(t+1) &= \frac{1}{\gamma}[\mathbf{Z}(t) - \mathbf{W}(t+1)\boldsymbol{\Xi}(t+1)\mathbf{Z}(t)],\end{aligned}\tag{55}$$

where  $\mathbf{I}$  is a unit matrix, and  $\gamma$  is a forgetting factor. Similarly, the APSA method for Equation (52) can also be formulated as follows [30, 35]:

$$\begin{aligned}\widehat{\mathbf{w}}(t+1) &= \widehat{\mathbf{w}}(t) + \frac{\mu\mathbf{X}^T(t) \operatorname{sgn}(\Delta(t))}{\sqrt{\operatorname{sgn}(\Delta^T(t))\mathbf{X}(t)\mathbf{X}^T(t) \operatorname{sgn}(\Delta(t)) + \chi}}, \\ \mathbf{X}(t) &= [\boldsymbol{\Xi}(t); \boldsymbol{\Xi}(t-1); \dots; \boldsymbol{\Xi}(t-M+1)], \\ \mathbf{Y}(t) &= [\boldsymbol{\delta}(t); \boldsymbol{\delta}(t-1); \dots; \boldsymbol{\delta}(t-M+1)], \\ \Delta(t) &= \mathbf{Y}(t) - \mathbf{X}(t)\widehat{\mathbf{w}}(t),\end{aligned}\tag{56}$$

where  $\mu$  is a forgetting factor and  $\chi$  represents the regularization parameter, which should be a positive number.

The identification result for each computation may vary due to the influence of the random measurement noise of the system. Thus, a total of  $N = 10$  simulations are conducted for each algorithm. Moreover, the arithmetic mean value (AMV), absolute error (AE), mean absolute deviation (MAD), standard deviation (SD), and relative standard


 FIGURE 3: Designed input torque signal  $\tau = 50 \sin(3.5t)$  ( $0 \leq t \leq 1.8$  s).

 FIGURE 4: Response signals of rotation angle  $\delta\theta$  and angular velocity  $\delta\dot{\theta}$  at the working point.

deviation (RSD) of the payload mass parameter  $m_e$  are, respectively, defined as follows:

$$\begin{aligned}
 \text{AMV} &\triangleq \bar{\hat{m}}_e = \frac{1}{N} \sum_{i=1}^N \hat{m}_e^{(i)}, \\
 \text{AE} &= \hat{m}_e^{(i)} - m_e, \quad (i = 1, 2, \dots, N), \\
 \text{MAD} &= \frac{1}{N} \sum_{i=1}^N \left| \hat{m}_e^{(i)} - m_e \right|, \\
 \text{SD} &= \sqrt{\frac{1}{N} \sum_{i=1}^N \left( \hat{m}_e^{(i)} - \bar{\hat{m}}_e \right)^2}, \\
 \text{RSD} &= \frac{\text{SD}}{\text{AMV}} \times 100\%,
 \end{aligned} \tag{57}$$

where  $N$  is the simulation times, and the corresponding AMV, AE, MAD, SD, and RSD for the moment of inertia  $J_e$  are similarly defined. The identified payload parameters  $\{\hat{m}_e, \hat{J}_e\}$  and relevant AMV using the three approaches are shown in Table 4. The AEs of the payload parameters using the three methods (10 simulations) are also shown in Figure 9.

The AMV results in Table 4 demonstrate that the proposed method can estimate the payload parameters more accurately than the other approaches. In Figure 9, the circular lines for the three methods are used to distinguish the maximum error distribution of the payload parameters around the original value (namely, the center of the figure). Similarly, Figure 9 also indicates that the proposed method better approximates the original value of the system than the two other classical algorithms.

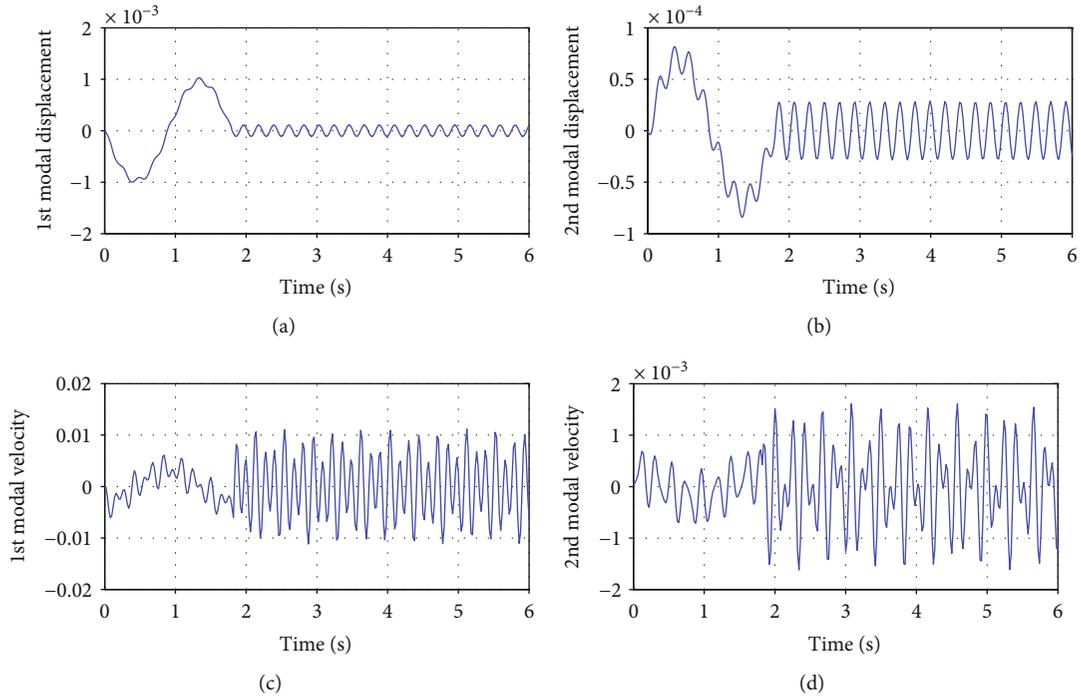
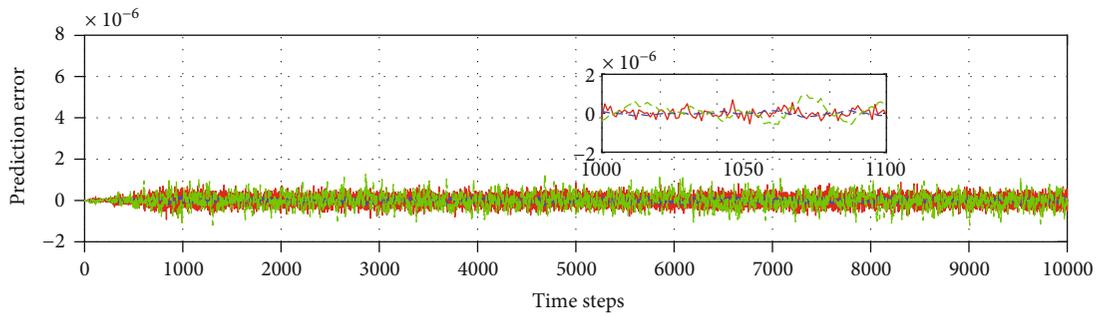
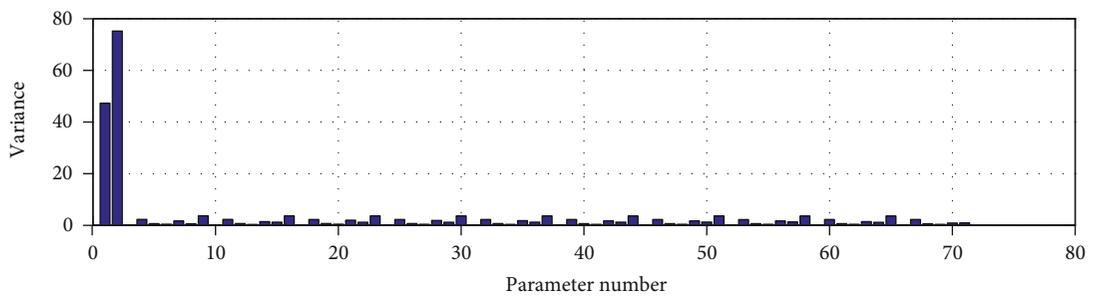


FIGURE 5: Response signals of modal displacements  $\delta\eta_i$  and modal velocities  $\delta\dot{\eta}_i$  at the working point ( $i = 1, 2$ ).



— Output 1  
 - - - Output 2  
 - - - Output 3

(a)



(b)

FIGURE 6: Prediction error for each step and computation variance of the OKID method.

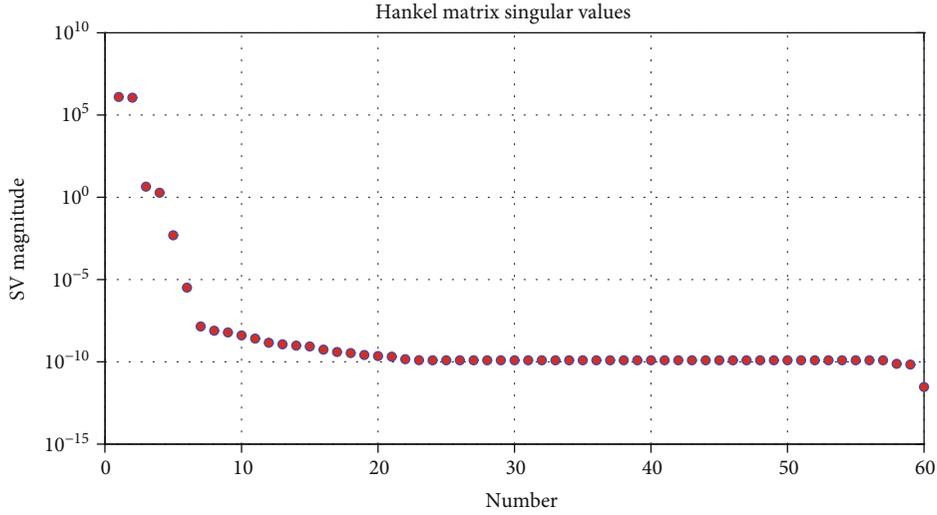


FIGURE 7: Singular values magnitude of the system by the OKID method.

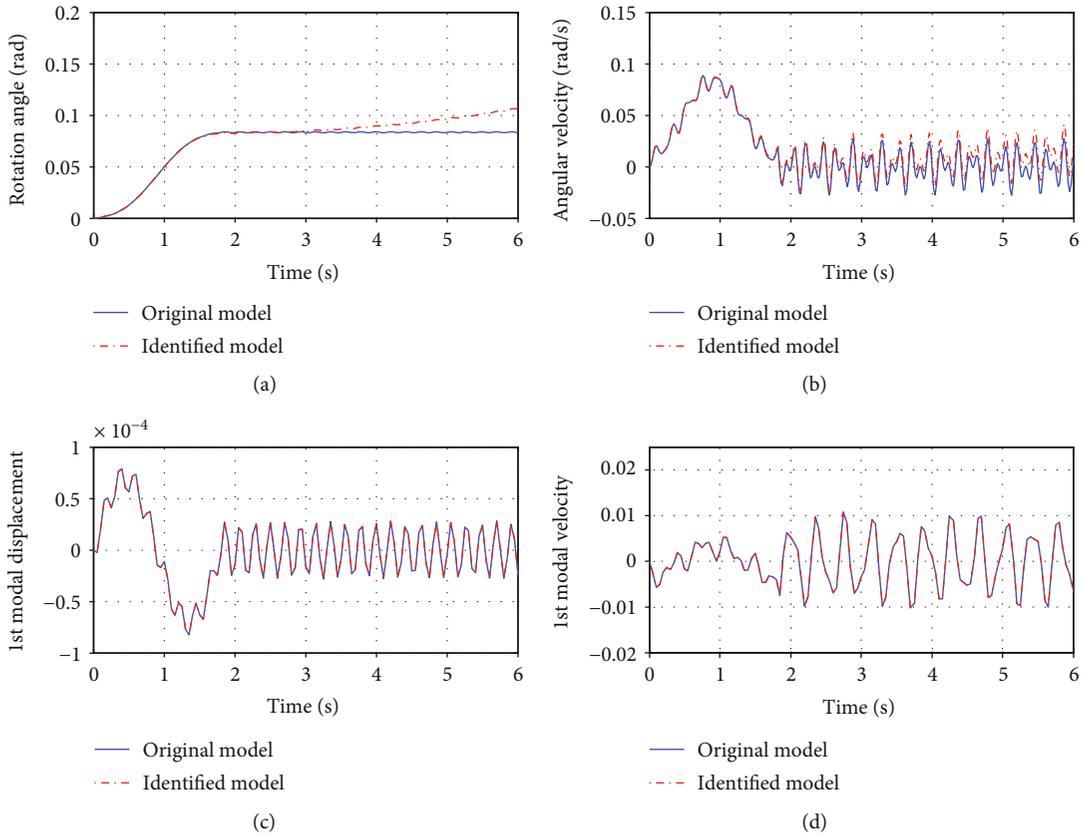


FIGURE 8: Test response signals of rotation angle  $\delta\theta$ , angular velocity  $\delta\dot{\theta}$ , 1st modal displacement  $\delta\eta_1$ , and 1st modal velocity  $\delta\dot{\eta}_1$  at the working point.

The MAD, SD, and RSD of the payload parameters for the three methods are shown in Figures 10 and 11. It can be seen that the proposed method has satisfactory identification accuracy and the corresponding RSD is less than 5%. Figures 9–11 demonstrate that and the computational errors by the proposed method are smaller than those by the classical RLS and APSA algorithms.

### 5. Conclusions

A payload parameter identification method of a flexible space manipulator system is proposed in this study. By linearizing the nonlinear dynamic equation of the manipulator at selected working points, the system state-space model and the corresponding eigenvalue parameters are, respectively,

TABLE 2: Average values of system complex eigenvalues  $\lambda_c$  between the original and identified system.

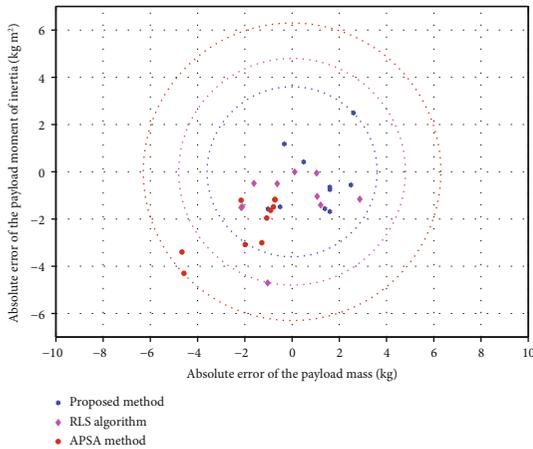
System order	Original values		Identified values	
	Real	Imag	Real	Imag
2nd	$2.5887e - 13$	$2.9452e + 01$	$1.9474e - 04$	$2.9452e + 01$
3rd	$2.8334e - 06$	$9.8878e + 02$	$-2.2964e - 07$	$9.8878e + 02$

TABLE 3: Average values of system complex mode shapes  $\psi$  between the original and identified system matrix  $A_c$ .

Real	Original values		Identified values	
	Imag	Real	Imag	
Mode 2				
$-1.9915e - 11$	$-3.3954e - 02$	$-2.4143e - 06$	$-3.2966e - 02$	
$4.7128e - 12$	$8.0349e - 03$	$1.6449e - 08$	$7.8011e - 03$	
$-1.1913e - 12$	$-2.0313e - 03$	$-3.0973e - 09$	$-1.9722e - 03$	
Mode 3				
$-2.8823e - 12$	$1.0113e - 03$	$-3.7741e - 11$	$8.8203e - 04$	
$1.6052e - 12$	$5.6450e - 04$	$2.9017e - 11$	$4.9232e - 04$	
$1.6084e - 13$	$-5.6848e - 05$	$1.3828e - 11$	$4.9579e - 05$	

TABLE 4: Identification results of the payload parameters (10 simulations).

Simulation no.	Mass $\hat{m}_c$ (kg)			Moment of inertia $\hat{J}_c$ (kg m <sup>2</sup> )		
	Proposed method	RLS algorithm	APSA method	Proposed method	RLS algorithm	APSA method
1	99.6703	99.2730	100.1103	201.1777	199.8258	199.4947
2	101.5996	99.2729	101.2102	198.3134	197.8104	198.8921
3	101.6014	98.7180	98.9655	199.2484	197.9906	195.2928
4	102.5959	99.2093	101.0607	202.4915	199.5080	198.9590
5	99.4926	97.8406	99.3710	199.5160	197.7941	199.4952
6	102.4911	95.3401	98.2425	199.4401	195.5997	198.7861
7	101.3892	95.4243	99.8859	198.4393	196.6997	198.0171
8	100.4916	98.0171	97.3875	200.4166	196.9188	199.8103
9	101.5955	98.9264	102.8673	199.3512	198.0419	198.8404
10	98.9814	99.0898	101.0377	198.4253	199.3663	199.9468
AMV	100.9909	98.1111	100.0139	199.6820	197.9555	198.7535

FIGURE 9: Absolute error (AE) of the payload parameters  $\hat{m}_c$  and  $\hat{J}_c$  (10 simulations for each algorithm).

determined using the OKID algorithm and the similarity transformation. Moreover, the payload parameters, namely, the mass and moment of inertia, can be derived by solving an LS problem based on the complex modal analysis of the system. The payload parameters obtained in the simulations by the proposed method are compared with those by the classical RLS and APSA algorithms. The results demonstrate that the proposed method can successfully be used to identify the manipulator payload parameters.

Certain problems related to the proposed method should be improved. (1) Only the planar situation is considered in this study, and the endpoint payload is simplified as a mass point. Therefore, the identification of the mass center is ignored. (2) In the identification procedures, the prior knowledge of model parameters, such as the coefficients in the mass matrix, is required. Otherwise, the accuracy of the solution will be affected. (3) Model order truncation is conducted in the

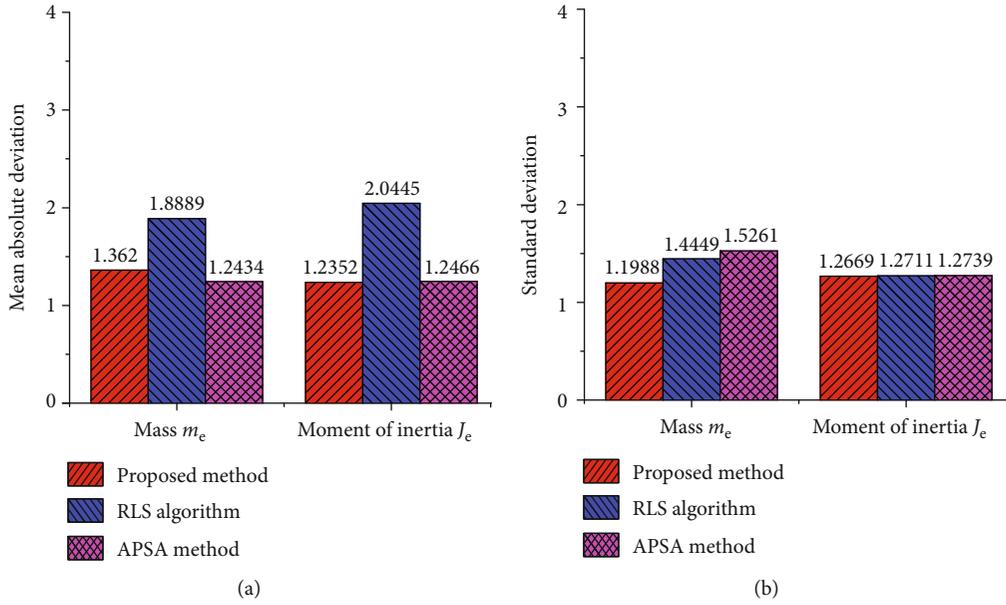


FIGURE 10: Computation results of the payload parameters  $\hat{m}_e$  and  $\hat{J}_e$  for the three methods: (a) mean absolute deviation and (b) standard deviation (10 simulations).

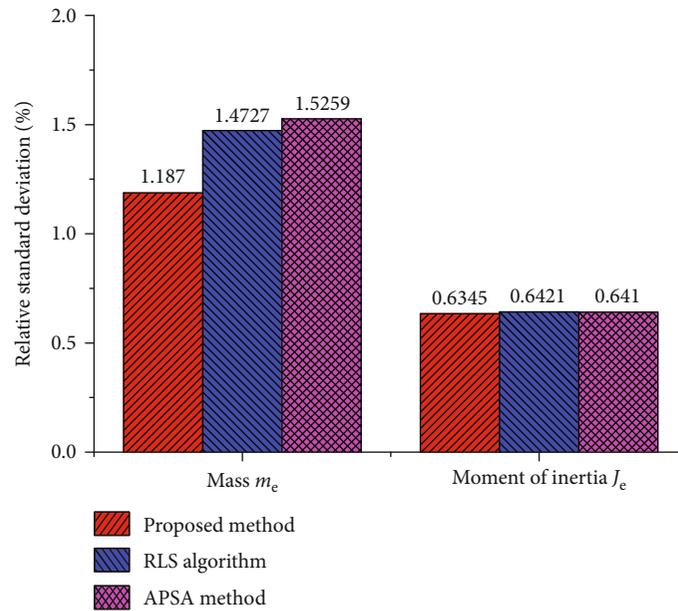


FIGURE 11: Relative standard deviation result of the payload parameters  $\hat{m}_e$  and  $\hat{J}_e$  for the three methods (10 simulations).

established manipulator system, and only the first two modal shapes of the link in Equation (1) are selected. However, if the number of selected vibration modes of the link is increased, then the establishment of the dynamic model would be considerably complicated. In future works, the extension of the proposed method to three-dimensional objects should be investigated, and a study will be conducted to reduce the effect of model truncation on the identification results.

## Appendix

### A. The Element Values of the Parameter Matrices in Nonlinear Dynamic Equation (9)

The detailed elements of mass matrix  $\mathbf{M}(\mathbf{q})$ , coupling coefficient matrix  $\mathbf{E}(\mathbf{q}, \dot{\mathbf{q}})$ , and stiffness matrix  $\mathbf{K}$  in nonlinear dynamics Equation (9) are expressed as follows, and other unspecified elements in the matrices are all zero.

$$\begin{aligned}
M_{11} &= \frac{\rho L^3}{3} + \rho L \eta_1^2 + \rho L \eta_2^2 + m_e L^2 \\
&\quad + 4m_e \eta_1^2 - 8m_e \eta_1 \eta_2 + 4m_e \eta_2^2 + J_e, \\
M_{12} &= M_{21} = 0.5688 \rho L^2 + 2m_e L + 2.7528 J_e, \\
M_{13} &= M_{31} = 0.0908 \rho L^2 - 2m_e L - 9.5614 J_e, \\
M_{22} &= \rho L + 4m_e + 7.5779 J_e, \\
M_{33} &= \rho L + 4m_e + 91.4204 J_e, \\
M_{23} &= M_{32} = -4m_e - 26.3206 J_e, \\
E_{11} &= 2\rho L \eta_1 \dot{\eta}_1 + 2\rho L \eta_2 \dot{\eta}_2 + 8m_e \eta_1 \dot{\eta}_1 \\
&\quad - 8m_e \dot{\eta}_1 \eta_2 - 8m_e \eta_1 \dot{\eta}_2 + 8m_e \eta_2 \dot{\eta}_2, \\
E_{21} &= \rho L \dot{\theta} \eta_1 + 4m_e \dot{\theta} \eta_1 - 4m_e \dot{\theta} \eta_2, \\
E_{31} &= \rho L \dot{\theta} \eta_2 + 4m_e \dot{\theta} \eta_2 - 4m_e \dot{\theta} \eta_1, \\
K_{22} &= \frac{12.3603 EI}{L^3}, \\
K_{33} &= \frac{485.4905 EI}{L^3}. \tag{A.1}
\end{aligned}$$

## B. The Element Values of the Parameter Matrices in Equation (52)

The detailed elements of matrices  $\Xi$  and  $\delta$  in Equation (52) which used to the least-squares method are expressed as follows.

$$\begin{aligned}
a_{11} &= \lambda_{c2}^2 \psi_{12} L^2 + 2\lambda_{c2}^2 \psi_{22} L - 2\lambda_{c2}^2 \psi_{32} L, \\
a_{12} &= \lambda_{c2}^2 \psi_{12} + 2.7528 \lambda_{c2}^2 \psi_{22} - 9.5614 \lambda_{c2}^2 \psi_{32}, \\
a_{21} &= 2\lambda_{c2}^2 \psi_{12} L + 4\lambda_{c2}^2 \psi_{22} - 4\lambda_{c2}^2 \psi_{32}, \\
a_{22} &= 2.7528 \lambda_{c2}^2 \psi_{12} + 7.5779 \lambda_{c2}^2 \psi_{22} - 26.3206 \lambda_{c2}^2 \psi_{32}, \\
a_{31} &= -2\lambda_{c2}^2 \psi_{12} L - 4\lambda_{c2}^2 \psi_{22} + 4\lambda_{c2}^2 \psi_{32}, \\
a_{32} &= -9.5614 \lambda_{c2}^2 \psi_{12} - 26.3206 \lambda_{c2}^2 \psi_{22} + 91.4204 \lambda_{c2}^2 \psi_{32}, \\
b_1 &= \frac{-\lambda_{c2}^2 \psi_{12} \rho L^3}{3} - 0.5688 \lambda_{c2}^2 \psi_{22} \rho L^2 \\
&\quad - 0.0908 \lambda_{c2}^2 \psi_{32} \rho L^2 - k_{11} \psi_{12}, \\
b_2 &= -0.5688 \lambda_{c2}^2 \psi_{12} \rho L^2 - \lambda_{c2}^2 \psi_{22} \rho L - k_{22} \psi_{22}, \\
b_3 &= -0.0908 \lambda_{c2}^2 \psi_{12} \rho L^2 - \lambda_{c2}^2 \psi_{32} \rho L - k_{33} \psi_{32}. \tag{B.2}
\end{aligned}$$

## Data Availability

Some or all data, models, or codes generated or used during the study are available from the corresponding author by request.

## Conflicts of Interest

The authors declare that there is no conflict of interest regarding the publication of this article.

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