Research Article

Design of a Broadband Inverted Conical Quadrifilar Helix Antenna

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1. Introduction

With the rapid development of GPS and SNS, antenna plays a more crucial role on the system performance. Moreover, cardioid-shaped radiation patterns with good circular polarization in a broad bandwidth are required tremendously for satellite communication and navigation systems [1].

Quadrifilar helix antenna (QHA), because of its small size, cardioid-shaped radiation pattern, and excellent circular polarization over a wide angular range, is a promising candidate for the GPS and SNS applications [2]. Nevertheless, the bandwidth of a conventional QHA operating under resonant modes is generally not very wide, which constrains its application. The bandwidth of a printed QHA is typically equal to 5 to 8%, and by using a variable arm width to obtain the tapered printed QHA its bandwidth can be increased to about 15% [3]. In recent years, several other techniques have been described in the literature to broaden the bandwidth of the QHA. In the literature [4], by introducing the conical helix geometry, an 18.5% bandwidth can be achieved. When the conical geometry and the tapering helices are combined, the antenna bandwidth reaches to more than 25% [5]. In the literature [6], a technique by adding a shorted parasitic helix is presented, and the folded printed QHA obtains 30% of impedance bandwidth in axial mode. Besides, there are also many techniques proposed for one QHA to cover the multiband [7–11].

In this document, the printed quadrifilar helical arms are wrapped in an inverted conical shape, called inverted conical printed quadrifilar helical antenna (ICPQHA), and are integrated directly above the ground plane of the feed network. By adjusting the dimensions of the inverted cone, the equivalent impedance of the antenna will be changed and matched, and the bandwidth of the proposed antenna is more than 39%. The theoretical and experimental results also have reasonably good agreements.

The reminder of this letter is organized as follows: Section 2 presents the antenna structure and design. Section 3 discusses the simulated and measured results, including bandwidth, radiation pattern, and circularly polarized ratio. Section 4 gives the conclusions.

2. Antenna Structure

A typical QHA consists of four helices equally spaced and excited with sequential phase variation, 0°, 90°, 180°, and
270°. In our work, an inverted conical geometry is introduced to enhance the bandwidth of the antenna, and this inverted conical geometry with the underneath ground is inspired by the discone antenna which shows characteristics of ultra-wideband [12]. The antenna is designed to operate in left-hand circular polarization (LHCP). The procedure for the antenna design can be divided into three steps. The first step is to design the radiating structure of the ICPQHA with four coaxial inputs. The second step is to design the feed network. The final step is to integrate the radiating structure with the feed network to form the complete model.

2.1. Antenna Configuration. Figure 1 shows the configuration of the proposed ICPQHA with its integrated-feed network underneath, operating at the center frequency of 3.85 GHz. The four helix-shaped radiating elements are printed onto a thin dielectric substrate of relative permittivity $\varepsilon_r = 2.94$ and of thickness $h = 0.254$ mm, wrapped around an inverted conical support made of cork wood and mounted on a small ground plane of the feed network. A square metal frame is set around the substrate for mounting and withstanding physical abuse, whose edge length is $W_g = 70$ mm.

The drawing function for generating the inverted conical helical arms is using the following parameters of the inverted conical helix: the initial radius $r_0$, pitch distance $d_p$, number of turns $N$, radius change per-turn $\Delta$, and the arm width $W$. One of the helical arms, starting from $+x$ axis in the $x$-$o$-$y$ plane, can be described by the following equations:

$$
\begin{align*}
    x &= \left( r_0 + \frac{\Delta \theta}{2\pi} \right) \cdot \cos \theta, \\
    y &= \left( r_0 + \frac{\Delta \theta}{2\pi} \right) \cdot \sin \theta, \\
    z &= \frac{\theta}{2\pi} d_p, \\
    0 &\leq \theta \leq 2N\pi.
\end{align*}
$$

The geometrical parameters of the final antenna are listed in Table 1.

Figures 2 and 3 show the effect of radius change per-turn $\Delta$ on the antenna performance. When $\Delta$ changes from positive to zero, negative, the antenna geometry changes from inverted cone to cylinder and cone, gradually. As $\Delta$ decreases from 3.5 mm to 0.5 mm, the $|S_{11}|$ as well as axial ratio (AR) performance is worse gradually, which shows that the inverted conical geometry has better performance and verifies the effectiveness of our design.

Impacts of the ground size on antenna radiation are also studied and shown in Figure 4. When the ground edge length $W_g$ increase from 70 mm to 90 mm, the antenna radiation has little variation and shows high stability.

2.2. Feed Network Configuration. The feeding network plays a critical role on ensuring the antenna’s performance. To excite
the four arms with constant amplitude as well as quadrant phase difference throughout the whole bandwidth, we design a wideband feed network in a shunt-fed format and use one 180° ring hybrid and two Wilkinson dividers to satisfy the requirement. The network is composed of two substrate layers with a ground plane in the middle. The 180° ring hybrid is printed on the lower surface of the bottom substrate layer, and the Wilkinson dividers are printed on the top surface of the upper substrate layer. The bottom and top layer of the network are connected with shorting pins. Figure 5 gives the geometry of the designed feed network. The designed feed network is centered at 3.85 GHz and uses the substrates with the permittivity of 2.55 and the thickness of 2 mm and 0.762 mm, respectively. And the top substrate is thicker because it also serves as a base for the inverted conical support made of cork wood.

Simulated magnitude response and phase differences of the feed network are shown in Figure 6. It is observed that the maximum difference of magnitude is within 0.2 dB in the band of 3 to 4.6 GHz. The simulated phase differences of the other three output ports, with respect to Port 2, are with sequential phase variation, 90°, 180°, and 270°.
3. Results and Discussion

The design works of the proposed ICPQHA are carried out with CST Microwave Studio software based on the finite integration technique (FIT) method. To verify the proposed antenna design, the prototype antenna was fabricated, as shown in Figure 7.

The measurement of S-parameter was carried out with vector network analyzer Agilent 8722ES, and the simulated and measured $|S_{11}|$ of the ICPQHA are shown in Figure 8 for comparisons. Good agreement between the simulated and measured results is observed, where the test antenna achieves an impedance bandwidth of over 39% for $|S_{11}| < -10$ dB, covering the frequency range from 3.1 GHz to 4.6 GHz.

The radiation patterns of the ICPQHA were measured in the far-field anechoic chamber. The simulated and measured radiation patterns at $\varphi = 0^\circ$ for 3.6 GHz and 4.4 GHz are depicted in Figure 9, which coincide with each other, especially for the copolarizations. And the polarization isolation of LHCP and RHCP is more than 20 dB. The measured and simulated AR patterns at $\varphi = 0^\circ$ for 3.6 GHz and 4.4 GHz are also plotted in Figure 10. It is observed that, though there is some discrepancy between the measured and simulated results due to the machining error, the AR is still below 3 dB when $\theta$ varies from $-60^\circ$ to $70^\circ$, showing that the antenna has a wide angular coverage, since the corresponding angular coverage for a typical microstrip antenna is only about $\pm 50^\circ$ [13].

Figure 11 plots the measured LHCP gain versus frequency. It is seen that a stable gain over 6 dBi is obtained from 3.2 GHz to 4.4 GHz and a little decrease in the high frequency but still over 5.4 dBi. The measured AR versus frequency...
is shown in Figure 12, where the ICPQHA has an effective 3 dB axial ratio bandwidth of greater than 31.5%, ranging from 3.35 GHz to 4.6 GHz.

4. Conclusion

In summary, a circularly polarized printed quadrifilar helix antenna with much wider frequency bandwidth is presented. The printed helical arms are wrapped in an inverted conical shape and a wideband feed network is integrated directly to excite high-performance LHCP radiation. The antenna impedance and axial ratio bandwidth values are more than 39% and 31.5%, respectively, which make the proposed antenna a promising candidate for satellite communication and navigation systems.

Conflict of Interests

The authors declare that there is no conflict of interests regarding the publication of this paper.
Figure 10: Measured and simulated AR. (a) 3.6 GHz and (b) 4.4 GHz.

Figure 11: Measured LHCP gain versus frequency.

Figure 12: Measured AR versus frequency.
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References


