Time-Delayed Feedback Tristable Stochastic Resonance Weak Fault Diagnosis Method and Its Application

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1. Introduction

In the early fault diagnosis of rotating machinery, the fault signal is often weak, and the working environment is mostly in the background of strong noise [1–3]. For example, in the early fault diagnosis of rolling bearings, the fault vibration signal itself is very weak, and it can easily be submerged in the surrounding noise, which causes interference in the identification of fault features [4]. Therefore, how to detect and extract weak fault signals from a background of strong noise is especially important for fault diagnosis of rotating machinery.

Pulses caused by rotating mechanical faults are weak and often submerged in strong background noise, which can affect the accuracy of fault detection. To solve this problem, we study the stochastic resonance phenomenon of a tristable potential system based on strong noise background and also investigate the influence of time-delayed feedback on this stochastic resonance model. The effects of time-delayed feedback strength on potential energy, steady-state probability density function, and signal-to-noise ratio (SNR) are discussed. The results show that stochastic resonance can be enhanced or suppressed by adjusting the delay time and feedback strength. Combined with bearing fault diagnosis simulation research and experimental verification evaluation, the proposed time-delayed feedback tristable stochastic resonance fault diagnosis method is more effective than the classical stochastic resonance method.
the classical stochastic resonance theory requires the input
signal frequency to be much less than 1. However, the
characteristic frequencies in mechanical systems in prac-
tical applications are usually in the range of tens, hundreds
of hertz, to several kilohertz, far exceeding the frequency
range required by classical stochastic resonance theory
[19]. To solve this problem, scholars use the frequency-
shifted and scaling transform method to solve the small-
parameter problem to achieve large-parameter stochastic
resonance [20, 21]. Subsequently, in order to further study
stochastic resonance, He et al. [22] proposed a new mul-
tiscale noise-tuning method that overcomes the limitations
of the small parameters of classical stochastic resonance
and improves the performance of classical stochastic res-
onance by using multiscale noise. Lei et al. [23] proposed an
adaptive stochastic resonance method that utilizes the
optimization ability of the ant colony algorithm to make
more accurate diagnosis of faults. Lu et al. [24] proposed
a sequence algorithm based on a multiscale noise-tuned
stochastic resonance method to achieve signal de-
modation, multiscale noise tuning, and bistable sto-
chastic resonance sequences. Li and Shi [25] proposed
a fault diagnosis method for rolling bearings based on
strong background noise. This method not only overcomes
the difficulty of selecting sensitive intrinsic mode function
but also enhances the weak fault feature by combining it
with adaptive stochastic resonance. Shi et al. [26] proposed
a weak signal detection method based on adaptive sto-
chastic resonance and analytical mode decomposi-
tion-ensemble empirical mode decomposition. This method
can not only enhance the signal amplitude but also effectively
detect the submerged weak multifrequency signal. Qiao
et al. [27] proposed an adaptive unsaturated bistable sto-
chastic resonance method that solved the problem of
inherent output saturation in classical stochastic
resonance.

Most of the above scholars’ research is directed toward
the stochastic resonance of the classical bistable model.
Classical bistable stochastic resonance has a single structure
and cannot form a richer potential structure. The potential
model cannot match the complex vibration signal, which
limits the enhancement ability of weak signals. In order to
further improve the extraction effect of stochastic resonance,
some research scholars proposed the tristable stochastic
resonance potential model. For example, Lu et al. [28]
proposed a new method for enhancing the periodic fault
signal of a rotating mechanical vibration using a tristable
mechanical vibration amplifier. Lu et al. [29] proposed a new
stochastic resonance method based on the Woods–Saxon
potential model and used it for fault diagnosis of bearings. In
summary, in most of the studies, although numerous tri-
stable stochastic resonances and other models were studied,
the effect of delay feedback on stochastic resonance was not
considered. However, some scholars have studied the delay
stochastic resonance. For example, Li et al. [30] proposed
a weak signal detection method based on a time-delayed
feedback monostable stochastic resonance system and
adaptive minimum entropy deconvolution; this method
achieves resonance detection of weak signals by selecting an
appropriate time delay, feedback strength, and rescale ratio
combined with a genetic algorithm. Lu et al. [31] proposed
a nonstationary weak signal detection method based on
a time-delayed feedback stochastic resonance model. This
method is more suitable for detecting signals with strong
nonlinear and nonstationary properties, as well as signals
subjected to severe multiscale noise interference. Aiming at
the problems existing in the above research, in this paper, we
propose a new time-delayed feedback tristable stochastic
resonance method and apply it to the diagnosis of weak
bearing faults. The periodic signal and Gaussian white noise
are introduced into the nonlinear tristable potential well,
and the delay term (including feedback strength and delay
time) is introduced. In the new potential system, through the
adjustment and optimization of the potential function pa-
rameters and the delay term parameters, the potential
function, the periodic signal, and the noise are optimally
coupled to obtain the optimal SNR. Compared with the
classical stochastic resonance method, the SNR is higher,
the fault characteristics are more apparent, and the output effect
is better than that of the classical stochastic resonance
method.

The rest of the paper is organized as follows: In Section 2,
the characteristics of the potential model are analyzed and
the influence of each parameter on the potential model is
discussed. In addition, a delay feedback term is introduced,
the Fokker–Planck equation is given, and the SNR is derived.
In Section 3, the process of fault detection and the simulation
signal of fault extraction are studied. In Section 4, the ex-
perimental verification of the bearing inner ring is con-
ducted to verify that the proposed method has higher SNR
for an actual measurement, and the reliability of the pro-
posed method is verified by using actual engineering data.
Finally, the conclusions are drawn.

2. Potential Model and Time-Delayed Feedback
Tristable Stochastic Resonance System

The process of SR detection of weak signals can be de-
scribed by particle motion in the potential well. When the
external drive is the periodic signal \( S(t) \), the particles
oscillate slightly on the side of the potential well. When
the noise \( \xi(t) \) is added to the system, the input signal
becomes \( S(t) + \xi(t) \), and the noise energy will be partially
transferred to the particle to overcome the barrier height
of the system. When the external drive is too large, the
particles in the system are too fast, called the over-
resonance. When the external drive is too small, the
particles cannot break through the barrier and can only
move in a potential well, which is called the under-
resonance. Since the periodic signal and noise are often
fixed, in order to ensure the best stochastic resonance
effect, it is necessary to adjust the potential model to make
the particles move stably between the potential wells.
Since the classical stochastic resonance is a bistable po-
tential, the potential model has a single structure and
cannot form a rich potential structure to match the
complex vibration signal. Therefore, in this paper, the
classical potential function is introduced [32]:

\[
S(t) = \left( \frac{\partial^2}{\partial x^2} + \frac{\partial}{\partial x} \xi(t) \right) x(t) + \xi(t)
\]
\[ U(x) = x^4(bx^2 - c)^2 + ax^2, \]  
(1)

where \( a, b, \) and \( c \) are potential model parameters and are nonzero real numbers, used to adjust the potential structure. In a nonlinear system, noise and periodic signals work together to produce a synergistic effect, and a stochastic resonance phenomenon occurs. As shown in Figure 1, the particles transition between the potential wells under the action of noise and periodic signals. The higher the position of the particles in the ordinate, the larger the particle energy. In Figure 1(a), it can be seen that as \( a \) increases, the potential barrier of the intermediate potential well increases, and the positions of the potential wells on both sides are raised. In the synergy of the periodic signal and the noise, it is easy to jump from the potential wells on both sides to the intermediate potential well. In Figure 1(b), it can be seen that as \( b \) increases, the depth of the wells on both sides decreases, and the transition between the wells is easier and less energy is required; in Figure 1(c), it can be seen that as \( c \) increases, the barrier of the intermediate potential well increases, and it is difficult for the particles to transition between the three potential wells, and the required energy will increase.

The classic stochastic resonance method does not take into account the influence of historical information on the system, and it focuses on a short-term memory system. To consider the influence of historical information on the system, a time-delayed memory system is as follows [33]:

\[ \frac{dx(t)}{dt} = x(t) - x^3(t) + \lambda x(t - \tau) + \sqrt{2D}\xi(t), \]  
(2)

where \( \lambda \) represents the feedback strength, \( \tau \) represents the delay time, and both are often used as the tempering coefficients of the system itself; \( D \) represents the noise intensity; and \( \xi(t) \) represents the Gaussian white noise and satisfies \( \langle \xi(t) \rangle = 0 \) and \( \langle \xi(t)\xi(t') \rangle = \delta(t - t') \). Equation (1) is substituted into equation (2). We can get a small delay bistable potential system driven by a weak periodic signal as follows [34]:

\[ \frac{dx(t)}{dt} = 6b^2x^5(t) - 8bcx^3(t) + 2(c^2 + a)x + \lambda x(t - \tau) + A\cos(\Omega t) + \sqrt{2D}\xi(t), \]  
(3)

where \( A \) represents the signal strength. Equation (2) is a non-Markov process, which is reduced to a Markov process by using the probability density method. The approximate time-delayed Fokker–Planck equation obtained is as follows [35]:

\[ \frac{\partial p(x, t)}{\partial t} = \frac{\partial}{\partial x} [h_{eff} p(x, t)] + \frac{\partial^2 p(x, t)}{\partial x^2}, \]  
(4)

where \( h_{eff} \) is the conditional average drift rate, given by

\[ h_{eff} = \int_b^a dx h(x, x_t) p(x_t, t - \tau | x, t), \]  
(5)

where the integral boundaries \( a \) and \( b \) are \( \pm \infty \), and \( p(x_t, t - \tau | x, t) \) is the zero-order approximate Markov transition probability density, whose formula is

\[ p(x_t, t - \tau | x, t) = \frac{1}{\sqrt{4\pi D\tau}} \exp\left(-\frac{(x_t - x - h(x)_T)^2}{4D\tau}\right), \]  
(6)

where \( h(x) = 6b^2x^5(t) - 8bcx^3(t) + 2(c^2 + a)x + \lambda x + A\cos(\Omega t) \).

Equation (6) is substituted into equation (5) to obtain

\[ h_{eff} = (1 + \lambda \tau)(6b^2x^5(t) - 8bcx^3(t) + 2(c^2 + a)x + \lambda x + A\cos(\Omega t) + \lambda (1 + \lambda \tau)A\cos(\Omega t)). \]  
(7)

Furthermore, obtaining an effective Langevin equation relative to equation (4), we have

\[ \frac{dx(t)}{dt} = 6b^2x^5(t) - 8bcx^3(t) + 2(c^2 + a)x + \lambda x + A\cos(\Omega t) + \sqrt{2D}\xi(t). \]  
(8)

Without considering the periodic signal and Gaussian noise, the equivalent time-delayed bistable potential corresponding to equation (8) can be derived as [36]

\[ U_{\xi}(x) = (1 + \lambda \tau)(x^4(bx^2 - c)^2 + ax^2) + \frac{1}{2}\lambda(1 + \lambda \tau)x^2. \]  
(9)

The effective potential function graph is shown in Figures 2(a) and 2(b).

The time-delayed feedback stationary probability density function is expressed as

\[ P_{\xi} = N \exp\left(-\frac{U_{\xi}}{D}\right), \]  
(10)

where \( N \) is a normalized constant, and the graph of \( P_{\xi} \) is shown in Figures 3(a) and 3(b).

Then, the power spectral density function can be derived as

\[ S(\omega) = \pi(1 + \lambda)M^2 [\delta(\Omega - \omega) + \delta(\Omega + \omega)] \]  

\[ + \left[1 - \frac{M^2}{2(N^2 + \Omega^2)}\right] \frac{2(1 + \lambda)N}{N^2 + \omega^2}, \]  
(11)

where \( \Omega \) is the angular frequency (\( \Omega = 2\pi f \), that is, the product of the number of vibrations per unit time and \( 2\pi \)); \( \omega \) is the output frequency [37]; \( S(\omega) \) is the output power spectral density and is obtained by the Wiener–Khintchine theorem, which is the Fourier transform of the correlation.
function of the Langevin equation, independent of the frequency \( \omega \); \( N = ((\sqrt{2} (1 + \lambda)(1 + \lambda \tau))/\pi) \exp(-(1 + \lambda)^2(1 + \lambda \tau)/{4D}); \) and \( M = (A/D)(1 + \lambda \tau) \sqrt{1 + \lambda N} \).

In equation (11), we are allowed to omit the last term \((M^2)/(2(N^2 + \Omega^2))\), so \( S(\omega) \) is only defined for positive \( \Omega \), and then equation (11) becomes [31]

\[
S(\omega) = S_1(\omega) + S_2(\omega),
\]

\[
S_1(\omega) = \frac{\pi(1 + \lambda)M^2}{2(N^2 + \omega^2)} \delta(\Omega - \omega),
\]

\[
S_2(\omega) = \frac{2(1 + \lambda)N}{N^2 + \omega^2},
\]

where \( S_1(\omega) \) is the output power spectral density and there is no influence of the periodic signal in \( S_2(\omega) \), so \( S_2(\omega) \) is the noise power spectral density.

The SNR is given by

\[
\text{SNR} = \frac{S_1(\omega)}{S_2(\omega)} = \frac{\sqrt{2}A^2}{4D^2}(1 + \lambda)^2(1 + \lambda \tau)^3 \cdot \exp\left[-\frac{(1 + \lambda)^2(1 + \lambda \tau)}{4D}\right].
\]

Taking \( dR/dD = 0 \), we can get \( D = (1/8)(1 + \lambda)^2(1 + \lambda \tau) = (1/2)\Delta U \). When \( D/\Delta U = 0.5 \), the SNR reaches a maximum.

Figure 2(a) shows that when other parameters are fixed and the feedback strength is too small, as shown by \( \lambda = -0.6 \), it is easier for the particles to jump from the potential wells on both sides to the middle potential well, and the required energy is low. However, the barrier value at this time is too high, which is not conducive to the transition of particles from the intermediate potential well to the potential wells on both sides, and the required energy is high. The excessive feedback strength is shown in the figure. When \( \lambda = 0.6 \), the

![Figure 1: Relationship between the three steady-state potential functions and \( x \). Influence of (a) \( a \), (b) \( b \), and (c) \( c \) on the potential model.](image-url)
minimum value of the potential wells on both sides is too low, which is also not conducive to the transition of the particles. It takes considerable energy for a particle to jump to the intermediate well. Figure 2(b) shows the effect of the delay time $\tau$ on the potential energy. As the delay time $\tau$ increases, the steepness of the potential wall of the potential well in the middle of the potential energy becomes lower, the position of the potential wells on both sides is reduced, and the particles can more easily transition among the three potential wells, requiring less energy. The adjustment of potential energy is achieved by adjusting the time-delayed feedback strength. When the external driving force is small, the potential energy can be increased by adjusting the delay parameter so that the particles can easily realize the transition between the potential wells, and vice versa, thereby achieving the best stochastic resonance effect.

Figures 3(a) and 3(b) show the effect of delay strength and delay time on the probability density function, respectively. It can be seen in Figure 3(a) that as the intensity of the delay time increases, the peak of the probability density decreases, and the difference in value between the peaks is large. Figure 3(b) also shows that as the time extension increases, the probability density function peak decreases, but the difference in value between the peaks is small. Thus, comparing the two graphs, one can see that the feedback strength has a greater influence than the time extension on

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**Figure 2:** Potential energy and $x$ when the parameters are as follows: (a) $\lambda$ is taken as $-0.6, 0,$ and $0.6$ and $a = 0.1, b = 0.1, c = 1,$ and $\tau = 0$; (b) $\tau$ is taken as $0, 0.2,$ and $0.4$ and $a = 0.1, b = 0.1, c = 1,$ and $\lambda = 0.6$.

**Figure 3:** Relationship between the stationary probability density function and $x$ with the corresponding parameters $a = 0.4, b = 0.26, c = 0.45, A = 0,$ and $D = 0.15$ for (a) different delay strengths and (b) different time extensions.
the probability density function. The larger the steady-state density function value, the more stable the transition of the particles between the potential wells, thus ensuring better stochastic resonance.

Figure 4(a) shows the effect of the feedback strength $\lambda$ on the SNR. As shown in the figure, as the feedback strength $\lambda$ increases, the SNR gradually increases to its peak value and then decreases. Figure 4(b) shows the effect of the delay time $\tau$ on the SNR; that is, as the delay time $\tau$ increases, the SNR increases and then decreases, while the feedback strength $\lambda$ affects the SNR, and the value of the feedback strength $\lambda$ becomes larger, and the value of the SNR peak is lower. From Figures 4(a) and 4(b), the influence of feedback strength $\lambda$ and delay time $\tau$ on the SNR is obvious. At the same time, the effects of the two coefficients on the SNR are not synergistic but are mutually suppressed. Therefore, the optimal SNR of the time-delayed feedback tristable stochastic resonance system is at $\lambda$ and $\tau$ values that ensure better transition of the probability density function value, the more stable the transition of the particles between the potential wells, thus ensuring better stochastic resonance.

Output SNR is used as the evaluation index for evaluating the output effect of the signal. The larger the SNR, the better the suppression noise and the signal extraction effect.

3. Simulation of Time-Delayed Feedback Tristable Stochastic Resonance

3.1. Signal Processing of Time-Delayed Tristable Stochastic Resonance. Because stochastic resonance theory is implemented under adiabatic approximation conditions, the above derivation is limited to small-parameter limits. In other words, the required frequency is much less than 1 Hz [19], and the fault frequency in engineering practice is usually tens of hertz or even hundreds of thousands of hertz. Therefore, we first use the frequency-shifted and scaling transform method to make the tristable stochastic resonance meet the small-parameter requirements, and then we use the fourth-order Runge–Kutta method to solve equation (3) as follows [24]:

$$\begin{align*}
K_1 &= h[-U' (x[n]) + \lambda x [n - \text{round}(\tau \times f_s)] + S[n] + N[n]], \\
K_2 &= h[-U' (x[n] + \frac{K_1}{2}) + \lambda x [n - \text{round}(\tau \times f_s) + \frac{K_1}{2}] + S[n] + N[n]], \\
K_3 &= h[-U' (x[n] + \frac{K_2}{2}) + \lambda x [n - \text{round}(\tau \times f_s) + \frac{K_2}{2}] + S[n + 1] + N[n + 1]], \\
K_4 &= h[-U' (x[n] + K_3) + \lambda x [n - \text{round}(\tau \times f_s) + K_3] + S[n + 1] + N[n]], \\
x[n + 1] &= x[n] + \frac{K_1 + 2K_2 + 2K_3 + K_4}{6},
\end{align*}$$

where $f_s$ is the sampling frequency and $h$ is the calculation step.

In this paper, the ant colony algorithm is inspired by the behavior of ants searching for food in nature, and it is a group intelligent optimization algorithm. The ant colony algorithm is based on the study of the collective foraging behavior of real ant colonies in nature, simulating the real ant colony collaboration process. As a general stochastic optimization method, the ant colony algorithm [38] has been successfully applied to a series of combinatorial optimizations and achieved good results. The ant colony algorithm is used to optimize the parameters of the time-delayed feedback tristable stochastic resonance system. Output SNR is used as the evaluation index to evaluate the performance of the time-delayed feedback tristable stochastic resonance method for extracting weak fault signals. The expression is as follows:

$$\text{SNR} = 10 \log_{10} \frac{A_d^2}{\sum_{i=1}^{N/2} A_i^2 - A_d^2},$$

where $N$ is the signal length. $A_d^2$ and $\sum_{i=1}^{N/2} A_i^2 - A_d^2$ are the amplitudes corresponding to the drive frequency $f_d$ and the strongest interference frequency $f_n$, respectively. The driving frequency refers to the fundamental frequency of the periodic signal, and the strongest interference frequency refers to the strongest noise frequency. The larger the SNR, the more obvious the difference between signal and noise. The specific process is as follows:

1. **Signal Preprocessing.** Signal demodulation is used to detect the drive signal and the small-frequency-parameter limitation through frequency-shifted and scaling transform.

2. **Parameter Initialization.** The parameter-matching algorithm is initialized, and the evaluation function of stochastic resonance is selected.

3. **Parameter Optimization.** The structural parameters $a$, $b$, and $c$ and the time-delayed parameters $\lambda$ and $\tau$ are again optimized by the ant colony algorithm to
obtain the optimal parameter combination. The optimal range is [0, 10].

(4) Output Calculation. Equation (2) is substituted into equation (14) to give the output waveform, and the SNR is calculated by using equation (15).

(5) Signal Postprocessing. The optimal parameters are substituted into the model for calculation, the spectrum is obtained, and the fault features are identified and diagnosed.

3.2. Simulation of Time-Delayed Tristable Stochastic Resonance. To verify the effect of the time-delayed feedback tristable stochastic resonance weak fault diagnosis method, the effect of the proposed method is analyzed by simulating the bearing fault signal. The simulated fault signal is as follows:

\[ S(t) = A \sin(2\pi ft) \cdot \exp \left\{ -d[t - n(t)T_d] \right\}^2 \]  

where \( A = 1 \) is the signal amplitude; \( t \) is the sampling time; \( f = 1 \) kHz is the carrier frequency; \( d = 6f_s \) reflects the attenuation rate; \( f_s = 10 \) kHz is the sampling frequency; \( n(t) = \left\lfloor t/T_d \right\rfloor \) \((n(t) = \text{floor}(t/T_d))\) control pulse periods appear; \( T_d = 1/f_d = 0.014 \) is the pulse interval; \( f_d = 72 \) kHz is the driving frequency; and the sampling time is 0.5 s.

Figures 5(a) and 5(b) show the noiseless signal and the noisy signal, respectively. Figure 5(c) shows the spectrum of the Hilbert transform, and Figure 5(d) shows the envelope spectrum of the noisy signal. No information related to the driving frequency can be found in the spectrum or in the envelope spectrum because of noise interference. Therefore, by relying solely on the envelope signal, the identification and extraction of the fault signal cannot be achieved.

Figures 6(a) and 6(b) show the time-domain waveform and frequency spectrum of the time-delayed feedback tristable stochastic resonance method. Figures 6(c) and 6(d) show the time-domain waveform and frequency spectrum of the classical stochastic resonance method, respectively. It can be seen from Figures 6(a) and 6(c) that the pulse contour of the time-delayed feedback tristable stochastic resonance method is clearer. Meanwhile, the peak values of the proposed method and classical stochastic resonance method are 0.1896 and 0.03615. It can be seen that the proposed method better extracts the fault feature than the classical stochastic resonance method. In Figure 6(d), the noise is strong in the fault frequency band. The above analysis shows that the time-delayed feedback tristable stochastic resonance method has a greater effect on the weak fault signal than the classical stochastic resonance method.

4. Experimental Verification

4.1. Experimental Verification of Time-Delayed Tristable Stochastic Resonance. To verify the effectiveness of the time-delayed feedback tristable stochastic resonance method, the proposed method is applied to a fault characteristic frequency extraction experiment of slightly worn rolling bearings. Rolling bearings are an important part of rotating machinery and one of the more easily damaged parts. Therefore, a fault signal extraction experiment with slight wear of the inner ring of the rolling bearing is used to verify the effect of the proposed method. The test bench is a comprehensive experimental bench for mechanical equipment failure, as shown in Figure 7. The experimental data comprise a sampling frequency of 5120 Hz and a rotational speed of 1800 rpm. The bearing type used in the experiment is ER-16K, bearing diameter is 38.5 mm, and bearing contact angle \( \alpha = 0 \). According to the vibration analysis theory, the bearing inner ring fault characteristic frequency is 162.69 Hz. The characteristic frequency in the figure is interfered by strong noise, so the fault features cannot be identified and extracted in the spectrum and the envelope spectrum in Figure 8. The time-delayed feedback tristable stochastic resonance method is applied to the extraction of the fault characteristic frequency of the rolling bearing. The time-domain waveform and spectrum of the
output signal are shown in Figures 9(a) and 9(b), respectively. It can be seen from the spectrum that the maximum peak frequency is 162 Hz with a spectral peak of 0.251 and is close to the theoretical value of 162.69 Hz. In order to verify that the proposed method is superior to the traditional method, the experimental data used are applied to classical stochastic resonance and the time-domain waveform and spectrum of the output signal are shown in Figures 9(c) and 9(d). In Figure 9(c), the waveform profile is clearer. In Figure 9(d), the fault frequency is at 162 Hz and the spectral peak is 0.2038. Comparing the two methods, the proposed method in this paper is more effective. We define the difference between the frequency peak of the fault characteristic and the peak of the noise spectrum as the degree of recognition. The fault feature recognition degree of the proposed method is 0.20195, and the recognition degree of the classical bistable stochastic resonance method is 0.0495. Therefore, we can conclude that the proposed method is more effective.
4.2. Engineering Verification of Time-Delayed Tristable Stochastic Resonance. Based on the proposed method, the robust results are verified in the above experiment. The effect of the time-delayed tristable stochastic resonance in extracting the bearing weak fault signal is better than that of the classical stochastic resonance method, and the extracted fault features are more obvious. The noise interference is smaller. We apply the proposed method to the fault feature extraction experiment of the bearing inner ring of a steel mill to further verify the effectiveness of the proposed method. The experimental bearing is shown in Figure 10. The bearing inner ring fault characteristic frequency is 24 Hz, and the sampling frequency is 2560 Hz. The time-domain waveform, spectrum, and envelope spectrum of the original signal are shown in Figure 11. In the spectrum of Figure 11(b), it is difficult to find information related to the frequency of the
fault characteristic because of the presence of the noise frequency. In the envelope spectrum of Figure 11(c), the fault characteristic frequency can be seen, but at the same time there are other strong noise frequencies. Therefore, the extraction of the bearing fault characteristics in actual engineering cannot be realized by using the envelope spectrum alone. To this end, the proposed method is applied to the experimental data, and Figures 12(a) and 12(b) are obtained. In the spectrum of Figure 12(b), it can be seen that the peak frequency of the highest spectrum is 24.32 Hz, which is very close to 24 Hz, so the proposed method can effectively extract the bearing fault characteristics in engineering practice.

At the same time, the engineering experimental data taken are applied to the classical stochastic resonance method to obtain the time-domain waveform and spectrum, as shown in Figures 12(c) and 12(d), respectively. Comparing the two methods shows that the frequency of fault characteristics in the spectrum is the same, which verifies the accuracy of the proposed method. At the same time, the spectral peaks of the fault characteristics of the two methods are 0.07584 and 0.0217, respectively. The recognition degree of fault characteristics is 0.03936 and 0.00545, respectively; therefore, the fault characteristics extracted by the proposed method are more obvious. Moreover, the proposed method yields less in-band noise in the fault characteristic frequency. Therefore, the time-delayed feedback tristable stochastic resonance method is better when applied in engineering practice.

The above analysis shows that the time-delayed tristable stochastic resonance system can get better signal output. From physical analysis, the time-delayed tristable stochastic resonance system has three potential wells and adjusts the potential structure with three parameters. Compared with the classical bistable stochastic resonance, the time-delayed tristable potential model can obtain a richer potential structure and match the complex vibration signal to achieve better stochastic resonance effect. Moreover, introducing a delay term in the potential model, the potential model can change the external driving energy to obtain a stable particle transition between the potential wells and finally get the best signal-to-noise ratio, which is the best signal enhancement effect. Therefore, it can be concluded that the signal output is better than the classical bistable stochastic resonance due to the existence of the tristable model and the delay term.

5. Conclusion
In this paper, we have studied the time-delayed feedback tristable stochastic resonance system. A method for time-delayed feedback tristable stochastic resonance weak faults is proposed. The main conclusions are as follows:

(1) The optimal potential energy, steady-state probability density function, and SNR can be obtained by
Figure 11: Engineering bearing signal. (a) Time-domain waveform, (b) spectrum, and (c) envelope spectrum.

Figure 12: Engineering bearing signal. (a) Time-domain waveform of the time-delayed feedback tristable stochastic resonance method. (b) Spectrum of the time-delayed feedback tristable stochastic resonance method. (c) Time-domain waveform of the classical stochastic resonance method. (d) Spectrum of the classical stochastic resonance method.
adapting the time-delayed and feedback strength. In other words, the optimal stochastic resonance effect can be obtained by adjusting the delay term.

(2) A new time-delayed feedback tristable stochastic resonance model is established. Compared with the classical bistable stochastic resonance model, the proposed model can obtain a richer structure by adjusting the system parameters, thus achieving matching with complex vibration signals.

(3) A new weak fault diagnosis method was proposed. The proposed method can not only extract the simulation and experimental signals of bearing faults, but also effectively extract the fault signals in the project. Compared with the classical stochastic resonance method, the weak fault features extracted by the proposed method have a good recognition.

Data Availability

The experiment is carried out on the mechanical failure comprehensive simulation test platform as shown in Figure 8. The signal is acquired by a Zonic Block/618 instrument connected to a computer. The experimental data can be obtained by sending an email to onylzx@126.com.

Conflicts of Interest

The authors declare that they have no conflicts of interest.

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