

Research Article

Rhombus-Inscribed Circular Ring Fractal Array Antenna for Millimeter-Wave 5G Applications

Muhammad Usman Tahir, Umair Rafique , and Muhammad Mansoor Ahmed

Department of Electrical Engineering, Capital University of Science and Technology, Islamabad, Pakistan

Correspondence should be addressed to Umair Rafique; umairrafique.1987@gmail.com

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A 1×4 planar fractal array antenna is designed and presented for millimeter-wave 5G applications. The array's top side is made up of a rhombus-inscribed circular ring fractal patch radiator, while the backside is made up of a square notch-loaded partial ground plane. To achieve high gain, a 1×4 corporate feeding network is used to excite the array elements. The designed array has an overall size of $28 \times 17.75 \text{ mm}^2$. From simulations as well as experimental results, it is observed that the designed array offers a wide impedance bandwidth in the frequency range of 22.8–29.2 GHz. Furthermore, a peak gain of 10.7 dBi with a radiation efficiency of > 95% is observed in the operating bandwidth.

1. Introduction

Fifth-generation (5G) communication technology will bring unique features for consumers. It will provide access to users everywhere and will select the best performance among other technologies such as Wireless Fidelity (Wi-Fi) and Wireless Local Area Network (WLAN). The selection of the best performance will not only be based on the throughput but also on the most appropriate metrics depending on the nature of the service. Furthermore, 5G systems will support other services such as the Internet of Things (IoT) by providing an internet connection to a vast number of objects. In addition to this, 5G systems will also cover services such as vehicle-to-vehicle communication (V2V) and remote health services.

5G technology provides extremely high data rates with low latency. To achieve a high data rate, there is a need for large bandwidth, which can be achieved with the use of higher frequency bands [1]. For this purpose, millimeter-wave (mm-wave) frequency bands are the best solution to achieve high bandwidth along with a high data rate. The International Telecommunication Union (ITU) allocated a few possible frequency bands in the mm-wave spectra, such as 26 GHz, 28 GHz, 38 GHz, and 60 GHz [2]. Of these bands,

the 26 GHz (24.25–27.5 GHz) and 28 GHz (26.5–29.5 GHz) frequency bands receive the most importance worldwide because they enable the operators to achieve speed, reliability, latency, and capacity. However, the losses are high at these frequencies due to the absorption of oxygen molecules [3, 4]. To mitigate this problem, high-gain antennas are required, and one of the techniques to enhance gain is to design antenna arrays. In addition, the mm-wave communication systems require compact antenna designs. This demand can be fulfilled with the use of planar antennas because they are small in size and have low fabrication costs, and their integration is easy with portable devices. Furthermore, at mm-wave frequencies, the small wavelength provides an extra advantage for designing compact and efficient antennas for 5G communication systems. In addition, it is realistic to design antennas that are physically small but electrically large enough to perform efficiently. Previously, many researchers have presented different antenna array designs for 5G communication systems.

In [5], a design of a 1×8 antipodal Vivaldi antenna (AVA) array was presented for 5G applications. The authors described that the AVA array operated from 24.55 GHz to 28.5 GHz and offered a peak gain of 11.32 dBi. The same kind of AVA configuration was utilized in [6]. In this design, the

top and bottom radiation elements were constructed using two different tapered slot configurations. The top radiator had an exponential-shaped curve, while the bottom radiator, connected to the ground plane, had an elliptical-shaped tapered curve. Moreover, a defected ground structure (DGS) was used to enhance isolation between array elements. In [7], a broadband planar array was designed for 5G broadcast applications. The array element was composed of two dipole radiators and a power splitter based on a substrate-integrated cavity (SIC). The dipole elements were placed side by side on both sides of the SIC-based power splitter. Although the presented configuration provides a wide impedance bandwidth from 24 GHz to 29.5 GHz and a gain of 10 dBi, due to its complex nature, it can rarely be used in 5G communication devices. In [8], a dual-hexagonal-shaped 1×4 planar antenna array was presented for the 28 GHz frequency band. A broadband corporate feeding network was utilized to feed the array elements. Furthermore, conducting vias were placed around the array elements to achieve high isolation. The results showed that the presented array had an impedance bandwidth of 10 GHz in the frequency range of 25–35 GHz. The same kinds of configurations were presented in [9, 10]. In [9], a spiral-shaped radiator with three hexagonal-shaped parasitic elements placed at the backside of the radiators was used to achieve a wide impedance bandwidth in the frequency range of 23.76–42.15 GHz and a gain of 11.5 dBi, while in [10], a rhombus-shaped patch radiator with square-shaped parasitic elements was utilized to achieve a bandwidth from 26 GHz to 30.63 GHz.

A 1×4 series-fed elliptical slot-loaded circular patch array for 5G communication systems was presented in [11]. Due to the use of an elliptical slot within the circular patch, a dual-frequency response at 28 GHz and 38 GHz was achieved. Moreover, the gain values were noted to be 7.6 dBi and 7.21 dBi for both bands. In [12], a bow-tie-like dipole antenna array was presented for 5G applications. With its compact size, the array had an impedance bandwidth ranging from 23.41 GHz to 33.92 GHz. They also presented a snow-flake fractal antenna array in [13] for the 28 GHz frequency band that operates in the frequency range of 25.28–29.04 GHz and has a peak gain of 10.12 dBi. In [14], a 1×4 series-fed linear array was presented for mm-wave applications. The designed array consists of a microstrip feed line and a series of short-circuited stubs periodically placed on one edge of the microstrip feed line. The presented configuration offered an impedance bandwidth of 21.40% with a peak gain of 12.18 dBi. The authors also realized a 4×4 two-dimensional (2D) array and observed an impedance bandwidth of 19.15% and a gain of 18.30 dBi in the center frequencies. Although the designs presented in [12, 13] provide wide impedance bandwidth, they are large in size.

In [15, 16], hook-like and donut-shaped planar antenna arrays were presented for 5G technology. In [17], a square-framed T-shaped planar array was presented for 5G communication systems. The ground plane configuration was the same as presented in [15, 16]. The presented designs were small in size, but they had less impedance bandwidth. In [18], an eight-element nonuniform series-fed dipole array

was designed for mm-wave wideband applications. The presented dipole array had an extremely wide impedance bandwidth ranging from 23.3 GHz to 51 GHz, but it suffers due to low gain. Recently, some researchers designed antenna arrays using substrate-integrated waveguide (SIW) technology [19–21] and achieved wide bandwidth and high gain, but the complex nature of the designs restricts their use in compact communication devices.

In this paper, a 1×4 planar fractal antenna array is designed for 5G mm-wave applications. The main objective of the proposed array is to achieve high gain for 28 GHz 5G applications. The single element of the array has a rhombus-inscribed circular ring fractal shape, while a notch-loaded partial ground plane is placed on the bottom side to achieve wide impedance bandwidth. The presented results show that the antenna array operates in the frequency range of 22.8–29.2 GHz and has a peak gain of 10.7 dBi. Furthermore, the presented array design is simple in nature and can be fabricated using low fabrication techniques, e.g., chemical etching.

2. Array Design

2.1. Single Element. As shown in Figure 1(a), a rhombus-inscribed circular ring fractal patch fed using a 50Ω microstrip feed line is designed on the top side of a dielectric substrate, while the backside consists of a square notch-loaded partial ground plane (see Figure 1(b)). A low-loss dielectric substrate, Rogers RT/duroid 5880, having a dielectric constant of 2.2 is being utilized for the antenna design. The thickness of the substrate is chosen to be 0.51 mm. The proposed antenna has the following overall design parameters: $W_{SUB} = 7$, $L_{SUB} = 7.5$, $R_1 = 2.6$, $R_2 = 1.5$, $L_1 = 3.6$, $L_2 = 1.6$, $L_F = 2$, $W_F = 1.4$, $L_G = 1.6$, $G = 0.5$, and $S = 0.25$ (all dimensions are in mm).

The design evolution of a single antenna element consists of four major steps, as illustrated in Figure 2. The iteration-1 design consists of a circular ring patch radiator fed using a 50Ω microstrip feed line, while a notch-loaded partial ground is designed on the bottom side, as shown in Figure 2(a). The dimensions of the circular ring are calculated using the design expressions described in [22]. In iteration-2, a rhombus shape is designed within the circular ring without changing the other dimensions, as shown in Figure 2(b). In iteration-3 and iteration-4, another circular ring and rhombus are designed, as depicted in Figures 2(c) and 2(d).

Figure 3 depicts reflection coefficient (S_{11}) characteristics for each iteration presented above. It is observed from the result of Figure 3 that the iteration-1 design offered an impedance bandwidth of 4.44 GHz (25.98–30.42 GHz). With every step, the bandwidth of the antenna is improving. Although iteration-2 and -3 designs provide better impedance matching, they have less bandwidth compared to iteration-4 design. The observed impedance bandwidth of the proposed design is equal to 10.72 GHz (22.28–33 GHz).

Figure 4(a) shows the simulated radiation properties of a single antenna element. From the figure, one can observe that the proposed antenna exhibits a bidirectional radiation pattern for the yz -plane and a quasi-omnidirectional pattern

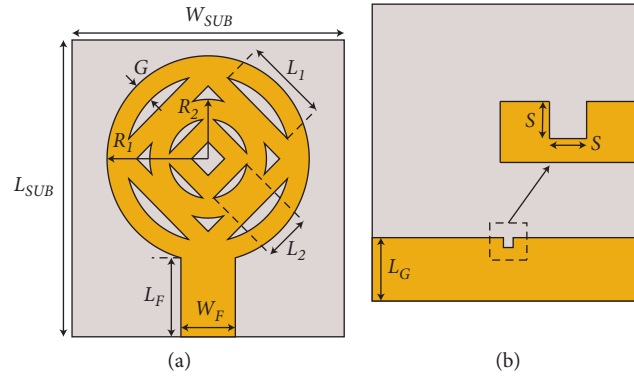


FIGURE 1: Schematic of the proposed fractal antenna: (a) front side; (b) back side.

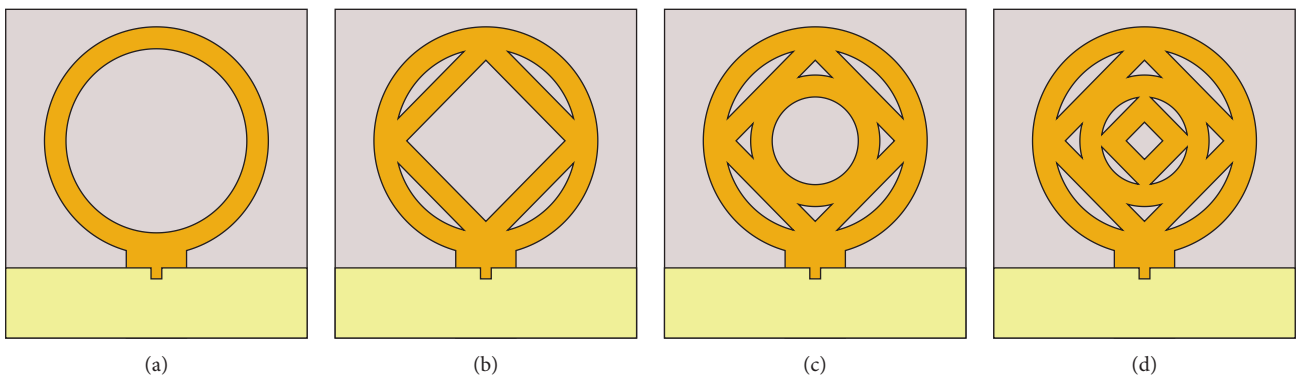


FIGURE 2: Design evolution of the proposed fractal antenna: (a) iteration-1; (b) iteration-2; (c) iteration-3; (d) iteration-4 (proposed design).

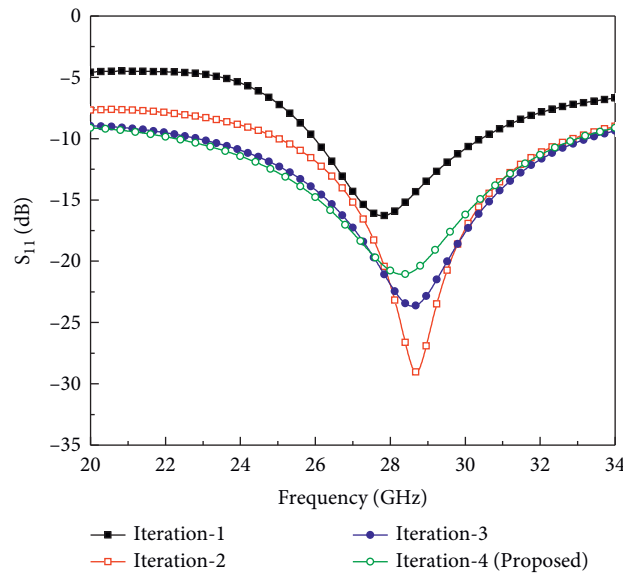


FIGURE 3: Simulated S_{11} for each iteration presented in Figure 2.

for the xz -plane. The simulated gain, radiation efficiency, and total efficiency of a single antenna element are illustrated in Figure 4(b). The gain of the antenna varies in the range of 3.5–4.5 dBi within the operating frequency range. Furthermore, the observed radiation and total efficiencies are $\geq 90\%$ and $> 80\%$, respectively.

2.2. Design of a 1×4 Antenna Array. To achieve high gain and narrow beamwidth, one of the requirements of 5G communication systems, a 1×4 linear array is designed, as shown in Figure 5. For array elements' feeding, a 1:4 conventional corporate feeding network is utilized, as shown in Figure 5. The corporate feeding network is designed using

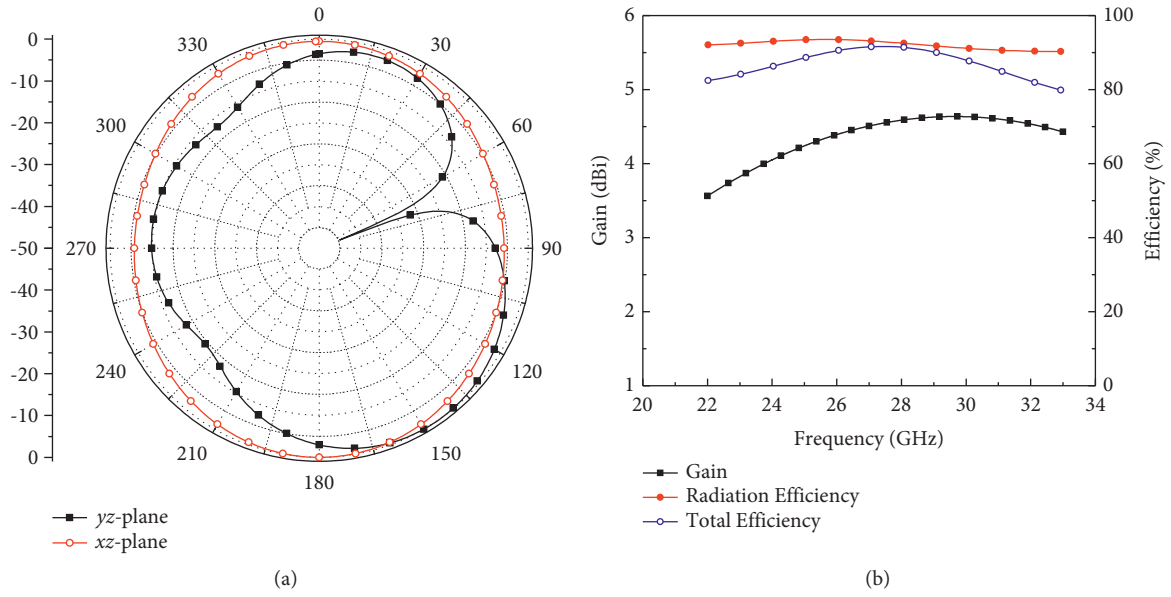


FIGURE 4: Simulated (a) radiation characteristics, (b) gain, radiation efficiency, and total efficiency of the proposed fractal antenna.

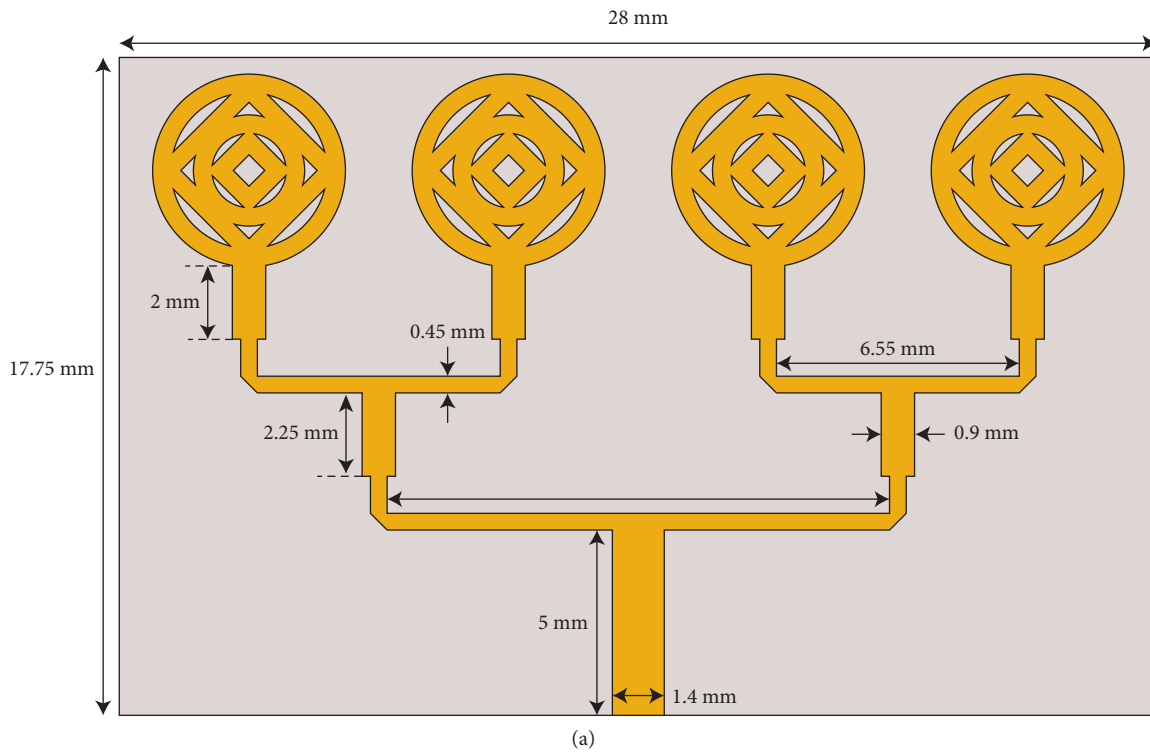


FIGURE 5: Continued.

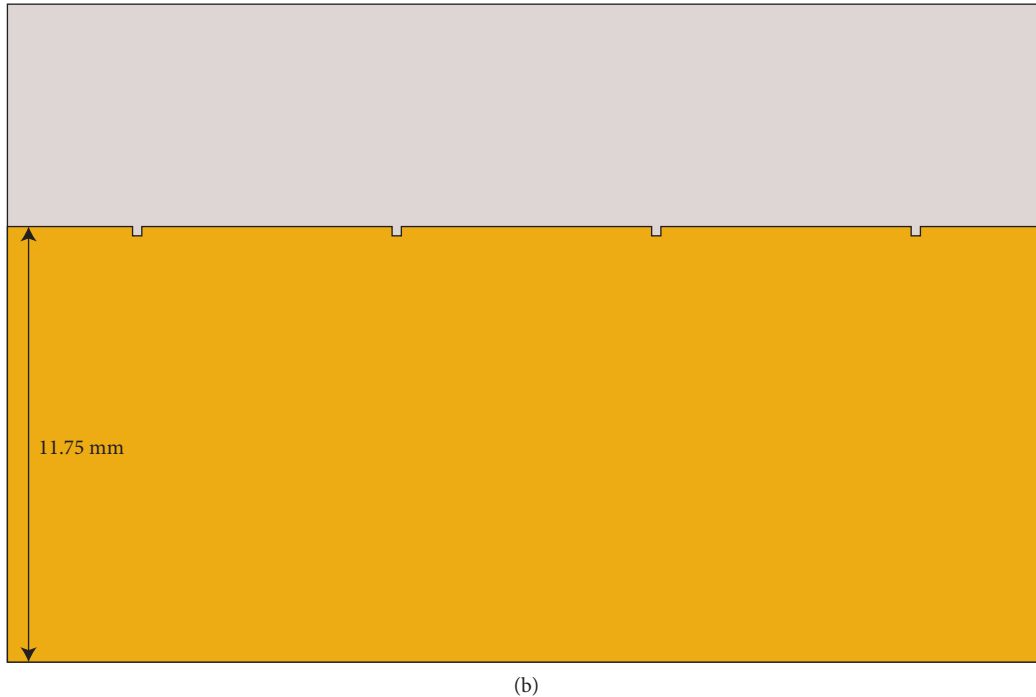


FIGURE 5: Design of the proposed 1×4 linear antenna array: (a) front view; (b) back view.

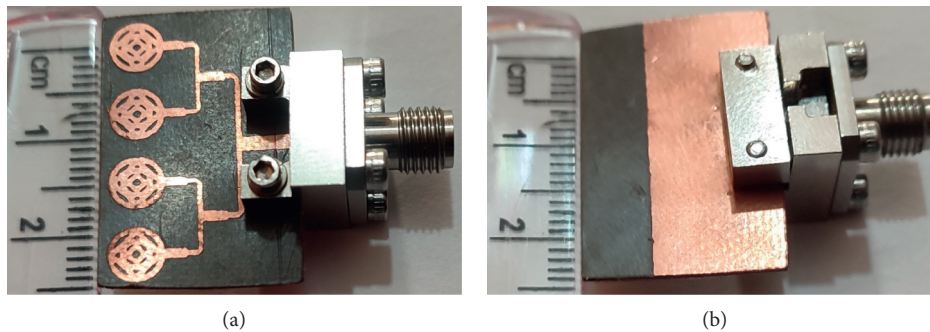


FIGURE 6: Fabricated prototype of the proposed array: (a) front view; (b) back view.

the design procedure presented in [17]. For better isolation, the distance between the array elements is chosen to be 7 mm, which is greater than $\lambda_0/2$ at 28 GHz, where λ_0 is the free-space wavelength.

The proposed four-element linear antenna array is designed and simulated in CST Microwave Studio and manufactured on a Rogers RT/duroid 5880 substrate with a thickness of 0.51 mm. The fabricated array prototype is shown in Figure 6. For measurement purposes, a 2.92 mm End Launch Screw-on type RF connector from Cinch Connectivity Solutions is used, as shown in Figure 6. For the validation of simulations, the fabricated prototype is measured using the Precision Network Analyzer (PNA) E8363C by Agilent Technologies.

The simulated and measured S_{11} characteristics of the antenna array are shown in Figure 7. From simulations, it is observed that, according to -10 dB bandwidth criteria, the designed array operates from 22.92 GHz to 29 GHz, while

the measurements show that the proposed array resonates in the frequency range of 22.8–29.2 GHz. Moreover, measured S_{11} is in good agreement with the simulated one.

For far-field measurements, the proposed array is characterized in an anechoic chamber using a conventional approach. The block representation of the measurement setup is shown in Figure 8. The proposed array is placed horizontally on a turntable on the opposite side of a horn antenna (reference antenna) having a frequency range of 15–40 GHz. To provide a reflection-free environment, the anechoic chamber walls are covered with radio frequency (RF) absorbers, as shown in Figure 8.

The gain of the proposed array is depicted in Figure 9(a), while the radiation efficiency is illustrated in Figure 9(b). Simulations show that the average gain of the proposed array is ≈ 10.3 dBi, while the measured gain has an average value of 10 dBi, as shown in Figure 9(a). Furthermore, as shown in Figure 9(b), the simulated and measured radiation efficiency

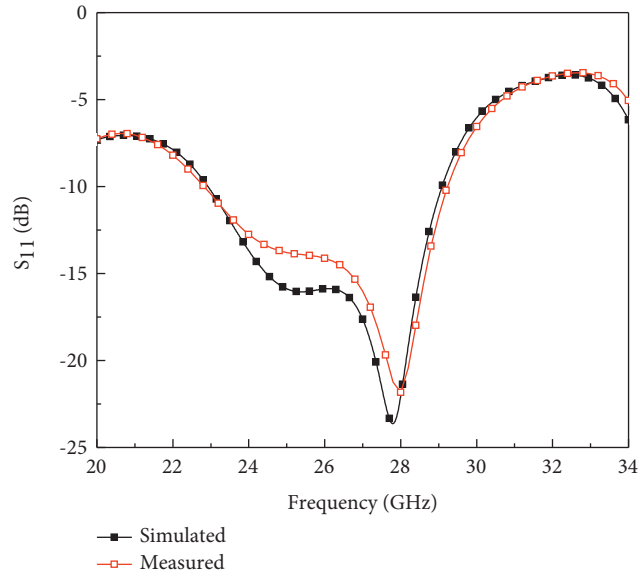


FIGURE 7: S_{11} characteristics of the proposed array.

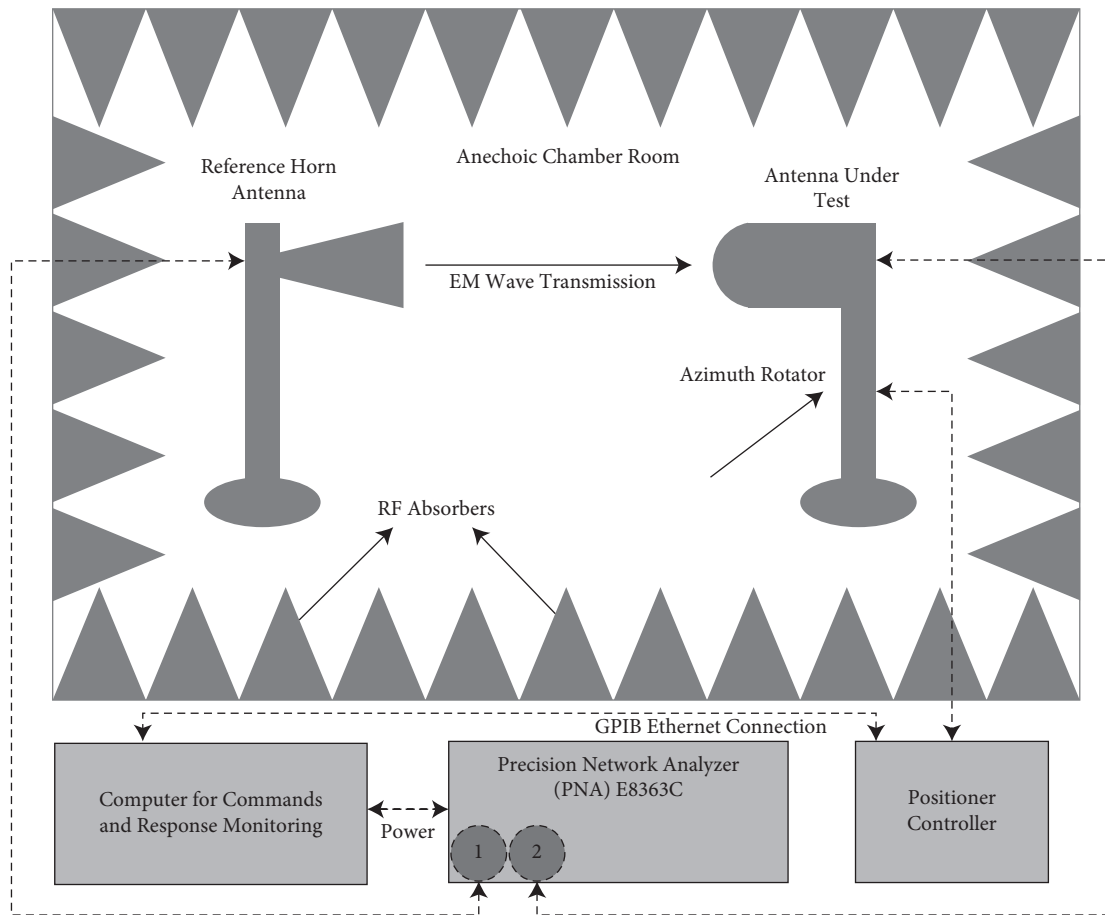


FIGURE 8: Block representation of the far-field measurement setup.

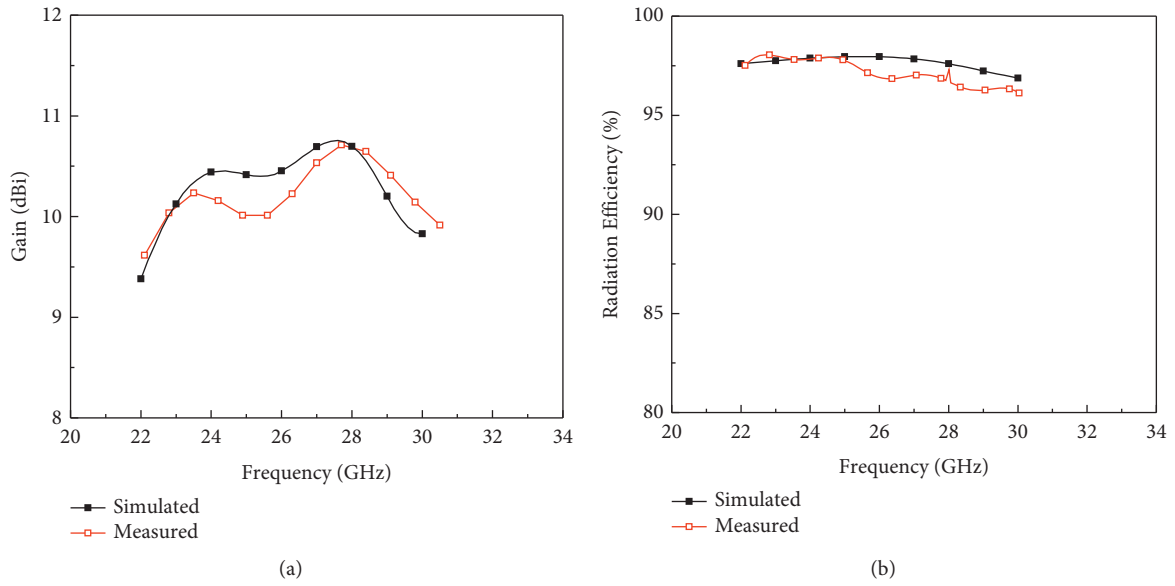


FIGURE 9: Simulated and measured (a) gain and (b) radiation efficiency of the proposed array.

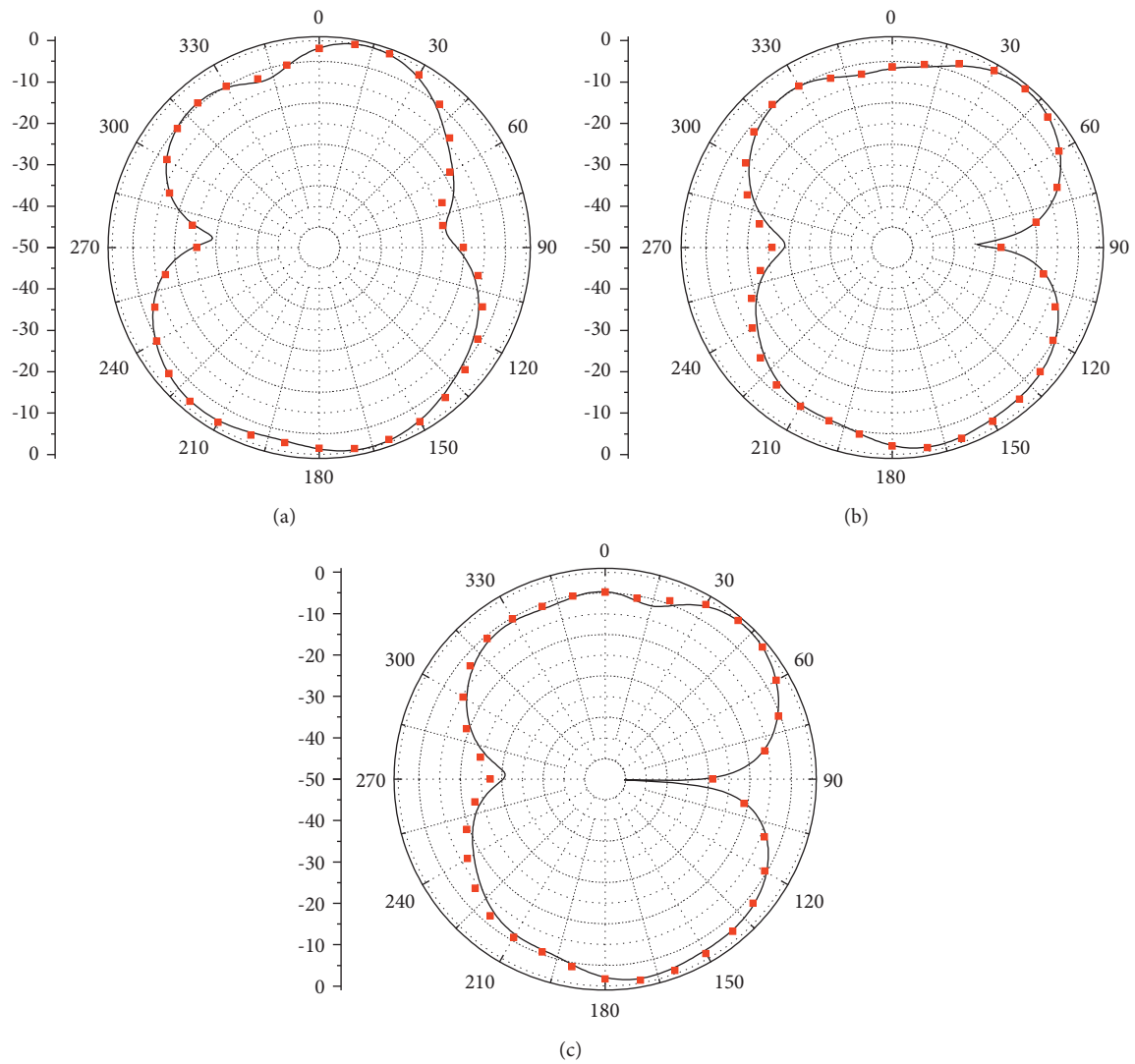


FIGURE 10: Simulated and measured radiation characteristics of the proposed array for the yz -plane at (a) 23 GHz, (b) 28 GHz, and (c) 29 GHz (— simulated and ■ measured).

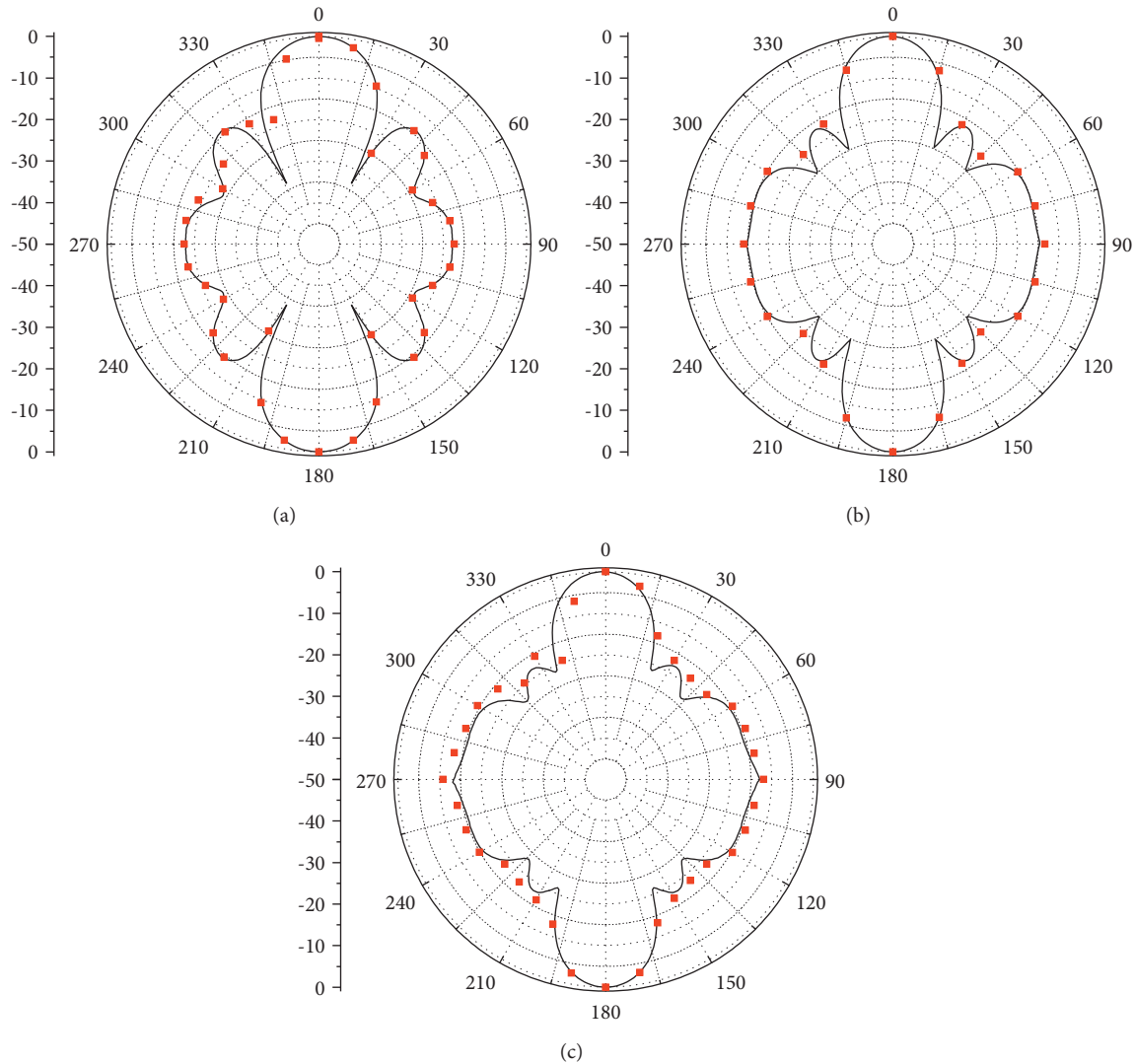


FIGURE 11: Simulated and measured radiation characteristics of the proposed array for the xz -plane at (a) 23 GHz, (b) 28 GHz, and (c) 29 GHz (— simulated and ■ measured).

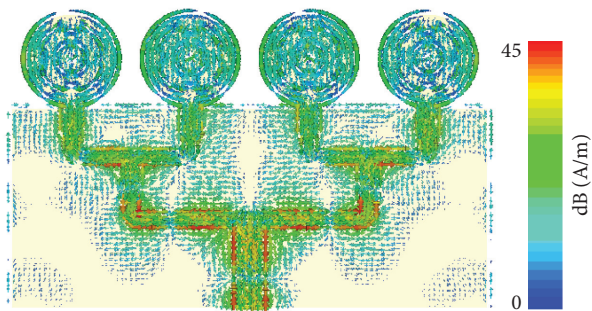


FIGURE 12: Surface current distribution for the proposed array.

is greater than 95% across the entire operating frequency band.

The radiation properties of the proposed array for the initial (23 GHz), middle (28 GHz), and end frequency (29 GHz) are shown in Figures 10 and 11. From Figure 10, for the yz -plane, one can observe that the proposed array exhibits a typical eight-shaped radiated pattern for the

frequency bands mentioned above. For the xz -plane, shown in Figure 11, highly directive dual-beam radiation properties are observed. In addition, a beamwidth of about 23° is observed for 23 GHz, while for 28 GHz and 29 GHz, it is $\approx 20^\circ$.

The simulated surface current distribution for the proposed array at 28 GHz is depicted in Figure 12. It is observed that the current is uniformly distributed along the feeding network and between the array elements. A uniform amplitude distribution of the surface current led to achieving high gain and smaller beamwidth, which can clearly be observed from the results of Figures 9(a) and 11.

3. Comparative Analysis

Table 1 presents a comparison between the proposed array and previously presented literature. The designs presented in [8–10, 12–14] offered large bandwidth or high gain, but they have a large size compared to the presented design. On the contrary, the designs presented in [15–17] are compact, but

TABLE 1: Comparison between the proposed array and previously published literature.

Reference	Dimensions (mm ²)	Array elements	Bandwidth (GHz)	Efficiency (%)	Gain (dBi)
[8]	45 × 20	4	10	85	12.15
[9]	40 × 15	4	15.42	83	11.5
[10]	40 × 19.22	4	4.63	85	11.24
[12]	37.6 × 14.3	4	10.51	90	10.7
[13]	32 × 12	4	3.76	80	10.12
[14]	26 × 40	4	4.92	—	18
[18]	34.1 × 10	8	27.7	—	10.8
[15]	26.9 × 18.5	4	2.4	90	10.3
[16]	22 × 20	4	3.7	90	10.2
[17]	24 × 18.5	4	3.7	94	11.5
This work	28 × 17.75	4	6.4	95	10.7

they suffer due to lower impedance bandwidth. Furthermore, the design presented in [18] offered extremely wide impedance bandwidth on the cost of increased array elements.

4. Conclusion

A 1 × 4 linear antenna array is designed for the 28 GHz frequency band. The single element of the array comprises a rhombus-inscribed circular ring fractal patch radiator and a notch-loaded partial ground plane. The designed antenna element demonstrates a wide impedance bandwidth from 22.28 GHz to 33 GHz. A 4-way corporate feeding network is designed for the antenna array. From simulations as well as measurements, it is observed that the designed array resonates well in the band of interest and offers a peak gain and radiation efficiency of 10.7 dBi and 95%, respectively. One can also conclude from the presented results that the designed antenna array is a potential candidate for future high-speed 5G communication systems.

Data Availability

No data were used to support this study.

Conflicts of Interest

The authors declare that they have no conflicts of interest.

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