

Research Article

Position Control of a Single-Rod Electro-Hydrostatic Actuator Experiencing a Leaky Piston Seal

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Single-rod cylinders are generally employed in electro-hydrostatic actuators (EHAs). A condition that is difficult to detect and could degrade the generally employed in electro-hydrostatic actuators (EHAs). A condition that is difficult to detect and could degrade the generally employed in electro-hydrostatic actuators (EHAs). A condition that is single-rod could degrade the generally encoded in electro-hydrostatic actuators (EHAs). A condition that is difficult to detect and could degrade the generally encoded in electro-hydrostatic actuators (EHAs). A condition that is electron in electron electr

1. Introduction

<An electro-hydrostatic actuator (EHA) is pump-controlled that has already been used for aircrafts [1], vehicles [2], and manipulators [3, 4]. In the literature, a double-rod cylinder is < configurations have more potential applications [5, 6]. ́ Single-rod EHAs can suffer from various faults. In partic-
en de se performance. In order to compensate for the adverse effects
groups, namely, active and passive FTC. In the former, the the latter, a robust controller is designed that is insensitive to < controllers is preferable due to its simple structure and easy application.

Quantitative feedback theory (QFT) is a robust linear controller design method. During the controller design process, performance specifications, parametric uncer tainties, and controller structure can be balanced [12, 13]. QFT controllers have been successfully applied to deal with internal leakage fault. Karpenko and Sepehri [14] developed
an active QFT-based FTC scheme, and they further designed passive QFT fault-tolerant controllers [15, 16]. All these controllers were implemented on valve-controlled systems. Ren et al. [17-20] synthesized QFT position and QFT actuating pressure controllers despite leakage. These controllers were developed for double-rod EHAs. Apart from Chen and Liu [21] has been proposed to deal with internal leakage. This system is not an EHA because directional developed a fractional-order PID fault-tolerant controller for a valve-controlled system. Moghaddam et al. [23] combined fractional-order PID controllers and a fuzzy inference system to accommodate internal leakage for a singlerod EHA. However, this active FTC strategy necessitates a fault detection algorithm. This active FTC strategy necessitates a fault detection algorithm. This active to the strategy necessitates a fault detection algorithm. The contribution of the strategy necessitates a fault detection algorithm. The contribution of the strategy necessitates and the the contribution of the strategy necessitates and the the strategy necessitates and the the necessitates and the the necessitates and the the necessitates and the the strategy necessitates and the the necessitates and the the necessitates and th

The remainder of this paper is organized as follows: mathematical model of this paper is organized as follows: mathematical model of this paper is organized as follows: mathematical model of this paper is organized in the set of the set of the two QFT controllers in simulations. Conclusions are provided in Section 5.

2. Modeling

The novel single-rod EHA circuit developed by Costa and Sepehri [6] is used for this structuit developed by Costa and Sepehri [6] is used for this structuit developed by Costa and Sepehri [6] is used for this structuit developed by Costa and Sepehri [6] is used for this structuit developed by Costa and Sepehri [6] is used for this structuit developed by Costa and Sepehri [6] is used for this structuit is structuit on the set is normal structuit on the set is normal structuit on the set is normal structuit of the set is normal structuit.
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From Figure 1, the flow equation of the main pump is

$$Q_1 = Q_2 = \omega_m V_d. \tag{1}$$

The flows out of and into the pump are represented by Q_1 and Q_2 , respectively; ω_m is the speed of the servo motor; the pump displacement is represented by V_d . The speed of the servo motor is [17].

$$\dot{\omega_m} = \tau_m \left(\left(-\omega_m \right) + \left(K_m u \right) \right). \tag{2}$$

Here, K_m represents the servomotor gain; t_m represents the time constant of the motor; u is used as input voltage to the motor. The actuator area ratio is

$$\alpha = \frac{A_b}{A_a},\tag{3}$$

 where A_a and A_b are the area of the piston at the cab side and
 the rot side, respectively. The load pressure is

$$p_L = p_a - \alpha p_b, \tag{4}$$

where p_a and p_b are the chamber pressures of the actuator. The continuity equations are



FIGURE 1: Schematic of the system developed in [6]. 1: Scrwomotor.
2: Bidirectional pump.3: Auxiliary pump. 4: Relief valve. 5: Ac-tuator. 6: Load mass. 7: Spring. 8: Tank. V1: One-directional flow control valve.

$$Q_{1} + Q_{ac} = Q_{a},$$

$$= A_{a}\dot{x_{p}} + \left(\frac{(V_{oa}) + (A_{a}x_{p})}{\beta_{e}}\right)\dot{p_{a}} + Q_{l},$$

$$Q_{bc} - Q_{2} = -Q_{b},$$
(5)

$$= -A_b \dot{x_p} + \left(\frac{(V_{ob}) - (A_b x_p)}{\beta_e}\right) \dot{p_b} - Q_l, \qquad (6)$$

where Q_{ac} and Q_{bc} are flows from the auxiliary circuit; Q_a and Q_b are the flows into and out of the actuator, respectively; x_p and $\dot{x_p}$ are the actuator position and velocity, respectively; β_e is the effective bulk modulus of the fluid; V_{oa} and V_{ob} are actuator chamber volumes at the two sides, respectively. The internal leakage Q_l is constructed as follows [17]:

$$Q_l = K_i (p_a - p_b). \tag{7}$$

In (7), K_i is the coefficient of the internal leakage. For single-rod actuators, the following assumption can be used [25,26].

$$\frac{(V_{oa}) + (A_a x_p)}{\beta_e} \approx \frac{(V_{ob}) - (A_b x_p)}{\beta_e} \approx \frac{(V_{oa}) + (V_{ob})}{2\beta_e} = C,$$
(8)

where C is the hydraulic compliance. The dynamic equation of the piston is

$$(m_{rod} + m_L)\ddot{x_p} = A_a p_L - f \dot{x_p} + F_L.$$
 (9)

In (9), $\vec{x_p}$ is actuator acceleration; m_{rod} is the piston and the rod mass; m_L is the load mass; f is the viscous damping coefficient; the load force F_L is



$$F_L = (m_{rod} + m_L)g - kx_p, \qquad (10) \qquad P_1$$

When the actuator is extending $(\dot{x_p} > 0)$, the auxiliary pump provides flow into one side of the actuator through V1 (Quadrants I and II in Figure 2). The following equation can be obtained:

$$Q_b - Q_l = \alpha (Q_a - Q_l). \tag{11}$$

From Equations (3) to (8) and (11), the following equations can be obtained:

$$(1 + \alpha^2)Q_a = (1 + \alpha^2)A_a\dot{x_p} + C\dot{p_L} + (1 + \alpha)K_ip_L,$$
 (12)

$$(1+\alpha^2)Q_b = \alpha(1+\alpha^2)A_a\dot{x_p} + C\alpha\dot{p_L} + (1+\alpha)K_ip_L.$$
 (13)

In Quadrant I, $p_L > 0$ and $Q_{ac} = 0$. Performing Laplace transformation of Equations (1), (2), (5), (9), (10), and (12), the following plant transfer function $P_1(s)$ can be obtained:

$$(s) = \frac{X_{p}(s)}{U(s)},$$

= $\frac{(1 + \alpha^{2})A_{a}V_{d}\tau_{m}K_{m}}{(s + \tau_{m})[(m_{rod} + m_{L})Cs^{3} + B_{1}s^{2} + A_{1}s + (1 + \alpha)kK_{i}]},$ (14)

where the constant A_1 and B_1 are

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$$A_{1} = kC + (1 + \alpha)fKi + (1 + \alpha^{2})A_{a}^{2},$$

$$B_{1} = fC + (1 + \alpha)(m_{rod} + m_{L})K_{i}.$$
(15)

$$P_{2}(s) = \frac{X_{p}(s)}{U(s)},$$

$$= \frac{(1+\alpha^{2})A_{a}V_{d}\tau_{m}K_{m}}{(s+\tau_{m})[\alpha(m_{rod}+m_{L})Cs^{3}+B_{2}s^{2}+A_{2}s+(1+\alpha)kK_{i}]},$$
(16)

where the constant A_2 and B_2 are

$$A_{2} = \alpha kC + (1+\alpha)fKi + \alpha (1+\alpha^{2})A_{a}^{2},$$

$$B_{2} = \alpha fC + (1+\alpha)(m_{rod} + m_{L})K_{i}.$$
(17)

When the actuator is retracting (x_p <0), the oil flows from one chamber of the actuator to the tank (Quadrants III and IV in Figure 2).

In Quadrant III, $p_L < 0$ and $Q_{bc} = 0$, Q_{ac} is [24].

$$Q_{ac} = -K_a p_a, \tag{18}$$

where K_a is the pressure sensitivity gain. Using Equations (1) to (10) together with (18), the plant model $P_3(s)$ can be obtained as follows:

$$P_{3}(s) = \frac{X_{p}(s)}{U(s)},$$

$$= \frac{[(1+\alpha)Cs + \alpha K_{a}]A_{a}V_{d}\tau_{m}K_{m}}{(s+\tau_{m})[(m_{rod}+m_{L})C^{2}s^{4} + C_{3}s^{3} + B_{3}s^{2} + A_{3}s + kK_{a}K_{i}]},$$
(19)

where the constant A_3 , B_3 , and C_3 are

$$\begin{aligned} A_{3} &= fK_{a}K_{i} + 2kCK_{i} + kCK_{a} + (1 - \alpha)^{2}A_{a}^{2}K_{i} + \alpha^{2}A_{a}^{2}K_{a}, \\ B_{3} &= (m_{rod} + m_{L})K_{a}K_{i} + 2fCK_{i} + fCK_{a} + kC^{2} + (1 + \alpha^{2})CA_{a}^{2}, \\ C_{3} &= 2(m_{rod} + m_{L})CK_{i} + (m_{rod} + m_{L})CK_{a} + fC^{2}. \end{aligned}$$

$$(20)$$

In Quadrant IV,
$$p_L > 0$$
 and $Q_{ac} = 0$, Q_{bc} is [24].

$$Q_{bc} = -K_b p_b, \tag{21}$$

where K_b is the pressure sensitivity gain. Using Equations (1) to (10) together with (21), the plant model $P_4(s)$ is

$$P_{4}(s) = \frac{X_{p}(s)}{U(s)},$$

$$= \frac{[(1+\alpha)Cs + K_{b}]A_{a}V_{d}\tau_{m}K_{m}}{(s+\tau_{m})[(m_{rod}+m_{L})C^{2}s^{4} + C_{4}s^{3} + B_{4}s^{2} + A_{4}s + kK_{b}K_{i}]},$$
(22)

where the constant A_4 , B_4 , and C_4 are

$$A_{4} = fK_{b}K_{i} + 2kCK_{i} + kCK_{b} + (1 - \alpha)^{2}A_{a}^{2}K_{i} + A_{a}^{2}K_{b},$$

$$B_{4} = (m_{rod} + m_{L})K_{b}K_{i} + 2fCK_{i} + fCK_{b} + kC^{2} + (1 + \alpha^{2})CA_{a}^{2},$$

$$C_{4} = 2(m_{rod} + m_{L})CK_{i} + (m_{rod} + m_{L})CK_{b} + fC^{2}.$$
(23)

The system includes the above four cases and is hereby the system model is expressed as $P(s) \in \{P_1(s), P_2(s), P_3(s), P_4(s)\}$. The internal leakage coefficient K_i and spring stiffness k change the plant type. Table 1 lists the parameter values [24] of the system. The minimum value of K_i is 0, which represents a healthy piston seal. The maximum value of K_i is prescribed to represent the most severe piston faulty

TABLE 1: Parameter values of the single-rod electro-hydrostatic actuator.

| Symbol | Value | |
|------------------|------------------------|---|
| | Nominal | Range |
| V _d | 8×10^{-6} | _ |
| τ _m | 3 | 2.3-4.0 |
| K _m | 5.8 | 5.6-6.0 |
| A _a | 3167×10^{-6} | _ |
| A | 0.75 | _ |
| K _i | _ | $0-2.4 	imes 10^{-11}$ |
| β _e | 689×10^{6} | $356 \times 10^{6} - 1030 \times 10^{6}$ |
| C | 3.46×10^{-12} | $2.20 \times 10^{-12} - 7.03 \times 10^{-12}$ |
| m _{rod} | 10 | 9–11 |
| m_L | _ | 0-300 |
| F | 900 | 600-1200 |
| Κ | _ | $0-130 \times 10^{3}$ |
| K_a, K_b | 4.65×10^{-10} | $2.08 \times 10^{-10} 4.65 \times 10^{-10}$ |

condition. Note that, the uncertainty range of m_L and k are also considered to ensure that both resistive and assistive load forces can be generated. These parameters are used in the simulations.

3. QFT Controller Design

Figure 3 shows the schematic of the control system [13]. As per prescribed specifications, a controller *G* and a prefilter *F* have to be synthesized despite uncertainties in the plant *P*.

3.1. Plant Templates. Parametric uncertainties of the plant (shown in Table 1) are captured by templates in the frequency domain on the Nichols chart. Template sizes are influenced by the effects of uncertainties on the plant. The templates of the plant P(s) for normal operation ($K_i = 0$) and the ones considering internal leakage ($K_i \ge 0$) are shown in Figures 4(a) and 4(b), respectively. Note that internal leakage increases templates sizes at low frequencies. It introduces phase variation with a maximum value of 90 degree and a magnitude variation. The larger templates make it hard to design the controller.

3.2. Prescribed Specifications

3.2.1. Tracking Specification. The uncertain plant P(s) is expressed as $P(s, \beta)$, where the vector $\beta = [\tau_m, K_m, f, C, m_{rod}, m_L, K_i, k, K_a, K_b]^T$. With reference to Figure 3, the transfer function of the closed-loop system $T(s, \beta)$ is

$$T(s,\beta) = F(s)\frac{G(s)P(s,\beta)}{1+G(s)P(s,\beta)},$$
(24)

where G(s) represents the QFT controller for normal operation $G_N(s)$ or the QFT fault-tolerant controller $G_{FTC}(s)$; F(s) is the prefilter designed for $G_N(s)$ or $G_{FTC}(s)$. The tracking requirement is shown as follows:



FIGURE 3: Schematic of the QFT control system.



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$$\begin{aligned} \left| T_{L}(j\omega) &= \frac{1}{\left((1/0.5s) + 1 \right) (s+1) \left((1/5s) + 1 \right) \left((1/30s) + 1 \right) \left((1/200s) + 1 \right)^{2}} \right| \\ &\leq \left| T(j\omega, \mathbf{\beta}) \right| \leq \left| T_{U}(j\omega) &= \frac{\left(1/0.55s \right) + 1}{\left((1/0.9s) + 1 \right) (s+1)^{2}} \right| \quad \forall \omega \in [0 \infty), \end{aligned}$$
(25)

where $T_U(s)$ is the upper tracking bound and $T_L(s)$ is the lower tracking bound. According to (25), the frequency responses of $T(s, \beta)$ should be within the above-given two tracking bounds.

3.2.2. Stability Specification. The stability specification is [13]

$$\left|\frac{G(j\omega)P(j\omega,\beta)}{1+G(j\omega)P(j\omega,\beta)}\right| \le 1.6 \ (4.1 \ dB) \quad \forall \boldsymbol{\omega} \in [0 \ \infty), \tag{26}$$

(26) ensures a gain margin of 4.22 dB and a phase margin of 36.42° [17, 18].

3.2.3. Sensitivity Specification. The following equation needs to be satisfied for disturbance rejection:

$$\left|\frac{1}{1+G(j\omega)P(j\omega,\beta)}\right| \le 1.8 \ (5.1 \ dB) \quad \forall \boldsymbol{\omega} \in [0 \ \infty).$$
(27)

3.3. Loop Shaping and Prefilter Design. A nominal plant $P(j\omega, \beta_0)$ is selected by using a set of parameters in $P(j\omega, \beta)$. It is then employed to calculate QFT bounds together with the above-prescribed specifications and plant templates on the Nichols chart. The controller is designed by shifting the nominal plant $P(j\omega, \beta_0)$ until the nominal loop transmission $L(j\omega, \beta_0) = G(j\omega) P(j\omega, \beta_0)$ satisfies QFT bounds et al. selected frequencies. The bounds are either open or closed. In order to satisfy these bounds, $L(j\omega, \beta_0)$ should be above the open bounds and outside closed bounds at the corresponding frequencies on the Nichols chart.

Figure 5(a) shows the QFT bounds and a suitable loop transmission for normal operation ($K_i = 0$). In order to satisfy open bounds, a small gain is employed in the controller. The next two poles are also added to satisfy closed bounds at high frequencies. The normal controller is shown in the following equation:

$$G_N(s) = \frac{50}{((1/6s) + 1)((1/6s) + 1)}.$$
 (28)

When the internal leakage is considered ($K_i \ge 0$), the QFT bounds and a suitable loop transmission are shown in



is figures is considering actuation (L) (jw, bo) (a) for normal operation and (b) considering actuator internal leakage.

Figure 5(b). An integrator is added in the fault-tolerant controller to make $L(j\omega, \beta_0)$ satisfy QFT bound requirements with a smaller controller bandwidth. Next, the open-loop gain is increased to meet open bounds. Finally, two zeros and two poles are used in the controller to satisfy closed bounds at intermediate frequencies and high frequencies, respectively. The designed fault-tolerant controller is shown in the following equation:

$$G_{FTC}(s) = \frac{80(s+1)((1/2s)+1)}{s((1/10s)+1)((1/20s)+1)}.$$
 (29)

By observing (28), an integrator, a high open-loop gain, and two zeros are needed in the fault-tolerant controller to cope with internal leakage fault. The ratio of controller gains $|G_{FTC}(s)|/|G_N(s)|$ is also calculated to further ascertain the price, as shown in Figure 6. It is seen that the integrator part in $G_{FTC}(s)$ introduces extra gain at low frequencies $(\omega < 1 \text{ rad/s})$ to remove static errors caused by leakage. At the intermediate-frequency band $(1 \text{ rad/s} \le \omega \le 10 \text{ rad/s})$, two zeros of $G_{FTC}(s)$ leads to over 8 dB ratio, that is required to satisfy its QFT bounds. Although the magnitude of $G_{FTC}(s)$ is much higher than that of $G_N(s)$ (ratio >25 dB) at high frequencies ($\omega > 10 \text{ rad/s}$), which indicates $G_{FTC}(s)$ is more susceptible to noise and unmodelled high-frequency dynamics, the prescribed specifications are still satisfied for both controllers.

Loop shaping just ensures the satisfactions of Equations (26), (27), and (30), therefore a prefilter was synthesized to meet (25). The prefilter can make closed-loop frequency responses within upper and lower QFT tracking bounds. The prefilter designed for normal operation and the one synthesized considering internal leakage are given by (31), respectively. Both prefilters have the same number of zeros and poles.

$$20 \log_{10} |T(j\omega,\beta)|_{max} - 20 \log_{10} |T(j\omega,\beta)|_{min} \le 20 \log_{10} |T_U(j\omega)| - 20 \log_{10} |T_L(j\omega)|,$$
(30)

$$F_N(s) = \frac{((1/2s) + 1)((1/4s) + 1)((1/150s) + 1)}{((1/0.5s) + 1)((1/30s) + 1)((1/500s) + 1)},$$
(31)

$$F_{FTC}(s) = \frac{((1/5s) + 1)((1/5s) + 1)((1/180s) + 1)}{((1/0.5s) + 1)(1/1.5s + 1)((1/15s) + 1)}.$$
(32)

4. Simulation Studies

The designed two QFT controllers were examined under normal operation (no leak) and leaky operation, respectively. Their ability to satisfy tracking bounds was shown in simulations. The ranges of parameters listed in Table 1 were also considered.

In the first test, the nominal system was operated under normal operation. A load of 300 kg and a spring of 130 kN/m

was chosen as the load force. Simulation results for the normal controller G_N and fault-tolerant controller G_{FTC} are shown in Figures 7 and 8, respectively. As is seen, the actuator position responses of the two controllers are within tracking bounds. In addition, when the quadrant switches, the control signal of G_{FTC} is more oscillatory than that of G_N . This is because the bandwidth of G_{FTC} is higher, making it more sensitive to the changes and disturbances of the system.



is figure 7: Simulation responses of QFT controller G_N and F_N to a 100-mm step input for actuator working under normal (no leak) condition, pushing against a 130 kN/m spring and moving a load mass of 300 kg. (a) Position. (b) Control signal. (c) Position error.



is figures est figures est de la controller G_{FTC} and F_{FTC} to a 100-mm step input for actuator working under normal (no leak) condition, pushing against a 130 kN/m spring and moving a load mass of 300 kg. (a) Position. (b) Control signal. (c) Position error.





in FIGURE 10: Simulation responses of QFT controller *G_{FTC}* and *F_{FTC}* to a 100-mm square-wave input for actuator working against a 130 kN/m spring and moving a load mass of 300 kg in the presence of increasing leakage flow. (a) Position. (b) Control signal. (c) Position error. (d) Leakage flow.

Next, internal leakage was gradually introduced (K_i increases from 0 to its maximum value) to evaluate the performance of the two controllers in leaky operation. The responses of G_N and G_{FTC} to a 100-mm square-wave input are shown in Figures 9 and 10, respectively. It is seen that the steady-state error of G_N increases with leakage and finally the tracking specification is violated. On the other hand, the position response of G_{FTC} satisfies tracking bounds, even when leakage increases to 4.4 L/min. **H** Finally, *G*_{*FTC*} was tested under various leakage levels Finally, *G*_{*FTC*} was tested under various leakage levels Finally, *G*_{*FTC*} was tested under various leakage levels is tested by the tested under tested by the tested (*K*), the tested by tested



is figure 11: Simulation responses of QFT controller G_{FTC} and F_{FTC} to various step inputs for normal operation as well as different leakage leakage leakage normal and 130 kg in free motion, against a 65-and 130-kN/m spring. (a) Position. (b) Control signal. (c) Position error. (d) Leakage flow.

5. Conclusions

A fault-tolerant controller was synthesized for a single-rod EHA. The controller required an integrator, a high openloop gain, and two zeros to compensate for leakage flow. Another QFT controller was also designed under normal operation (no leak). Simulation results demonstrated that the QFT fault-tolerant controller was capable of maintaining actuator responses within tracking bounds despite internal leakage up to 8.6 L/min. However, the QFT normal controller could not satisfy the prescribed specifications if internal leakage occurred.

Data Availability

The data used to support the findings of this study are available from the corresponding author upon request.

Conflicts of Interest

The authors declare that there are no conflicts of interest regarding the publication of this paper.

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